

# Active Stoppers for Decay Experiments

Valencia, Spain, March 24-26 2025



**Organizing Committee:** Nic Hubbard (GSI)  
Anabel Morales (IFIC, CSIC-UV) Shunji Nishimura (RIKEN)  
Tom Davinson (U. Edinburgh) Marta Poletini (GSI)  
Andrés Gadea (IFIC, CSIC-UV) Jelena Vesic (IJS)



# ADVANCES IN INTEGRATED FRONT-END CIRCUITS FOR SEMICONDUCTOR DETECTORS

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*INFN of Milano, Milano, Italy*

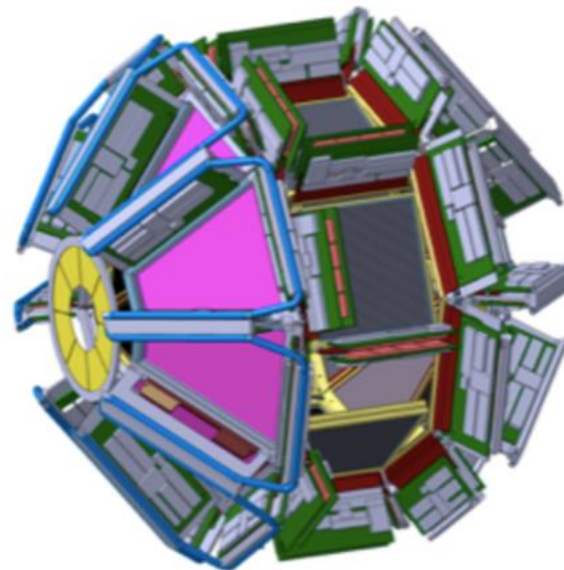


My background: integrated electronics for Particle spectroscopy with silicon detectors for research in nuclear physics

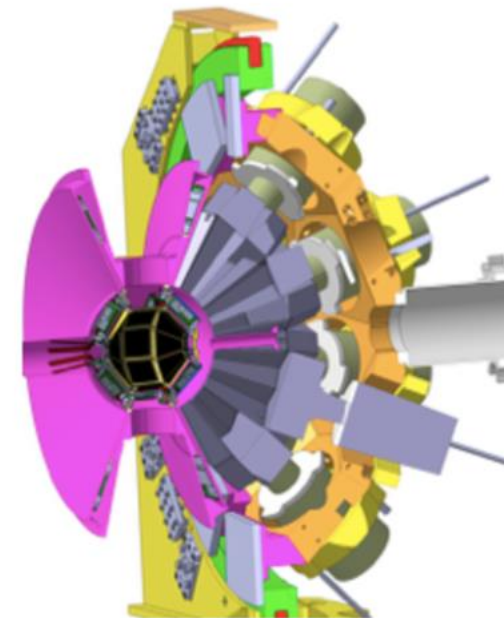


The goal: perform particle spectroscopy down to 0.5 MeV but also up to some hundreds of MeV

GRIT ARRAY



GRIT AND AGATA



0.5 MeV

500 MeV



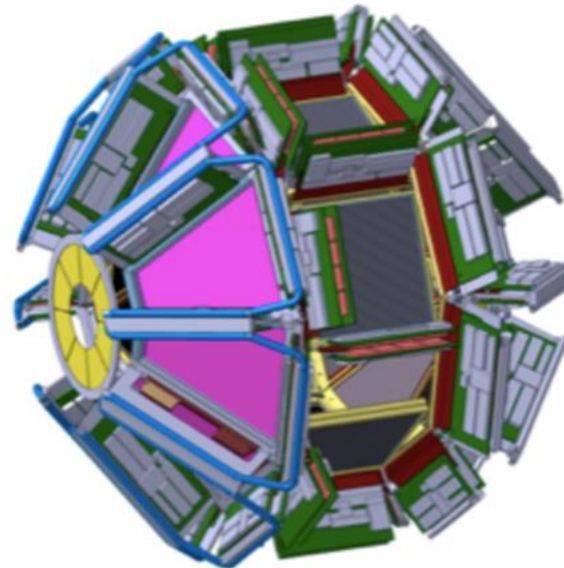


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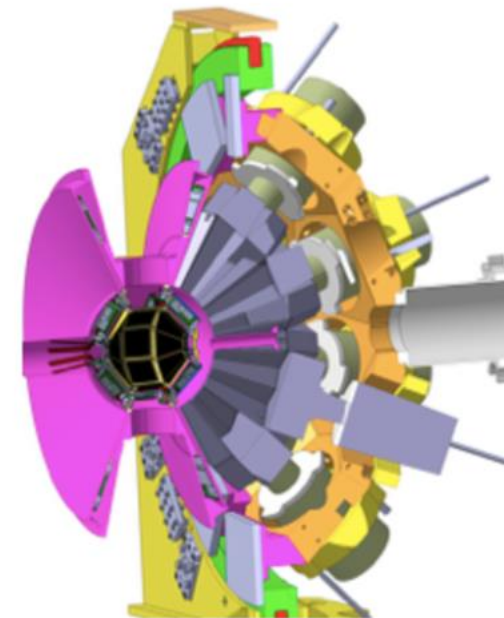
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Hybrid readout techniques

GRIT ARRAY



GRIT AND AGATA



0.5 MeV

600 MeV

Comfort zone



My background: integrated electronics for Particle spectroscopy with silicon detectors for research in nuclear physics



The goal: perform particle spectroscopy down to 0.5 MeV but also up to some hundreds of MeV



Hybrid readout techniques

Tens of keV  
(beta)

0.5 MeV

500 MeV

Several GeV  
(HI-implant)





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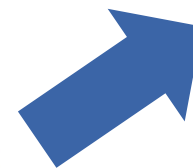
Hybrid readout techniques

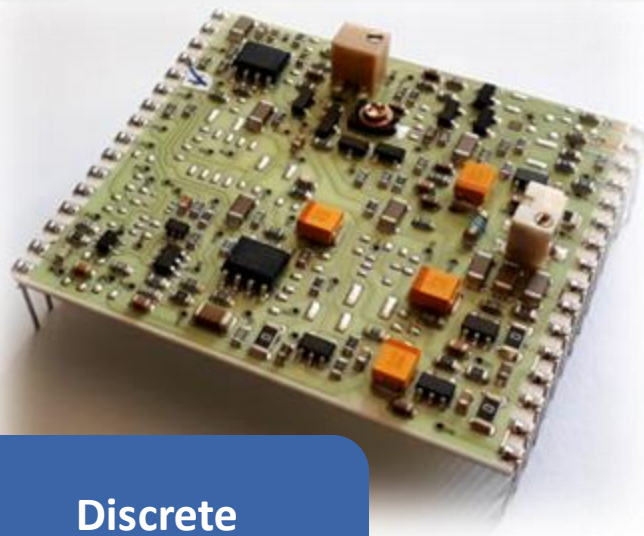
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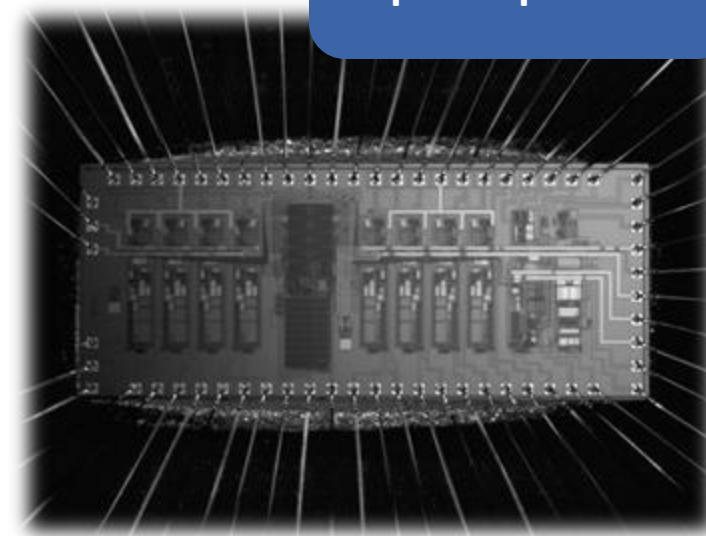


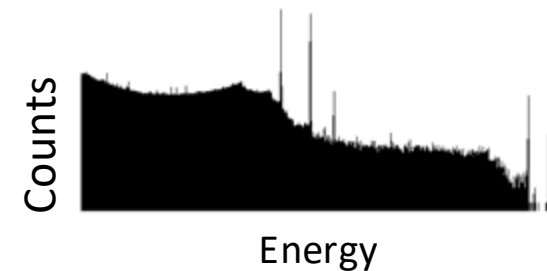
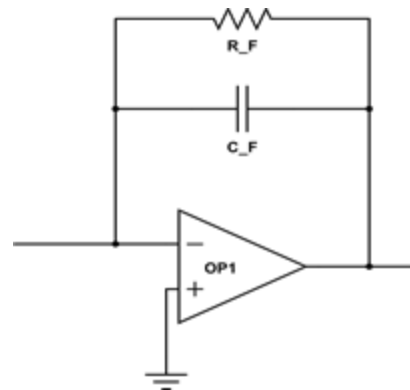
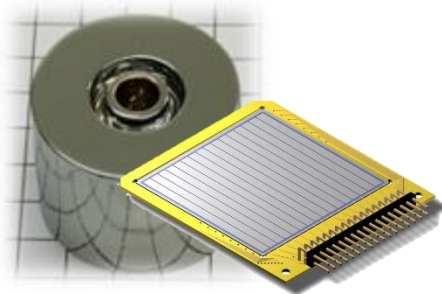
**Discrete preamplifiers**

- Not suitable in case of high channel density
- Components tolerant to higher bias voltages -> High dynamic range
- Higher power consumption
- Design flexibility

- Suitable in case of high channel density
- Components tolerant to lower bias voltages -> Low dynamic range
- Low power consumption
- Radio-purity

**Integrated preamplifiers**

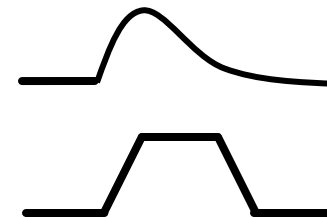




Solid-state  
detector

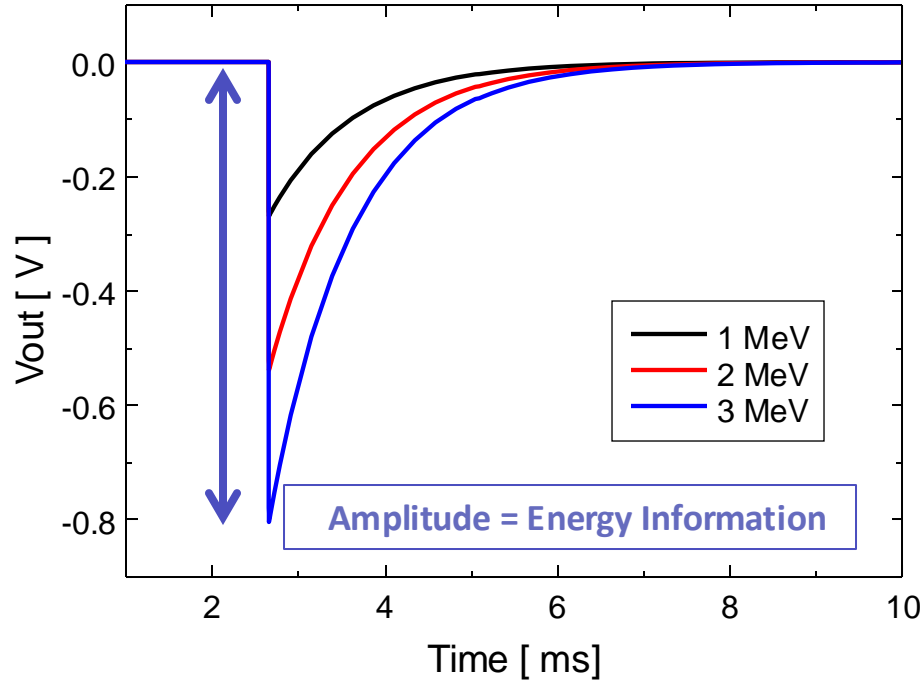
Charge-Sensitive  
Preamplifier

Amplifier and  
MCA or ADC with  
digital filtering

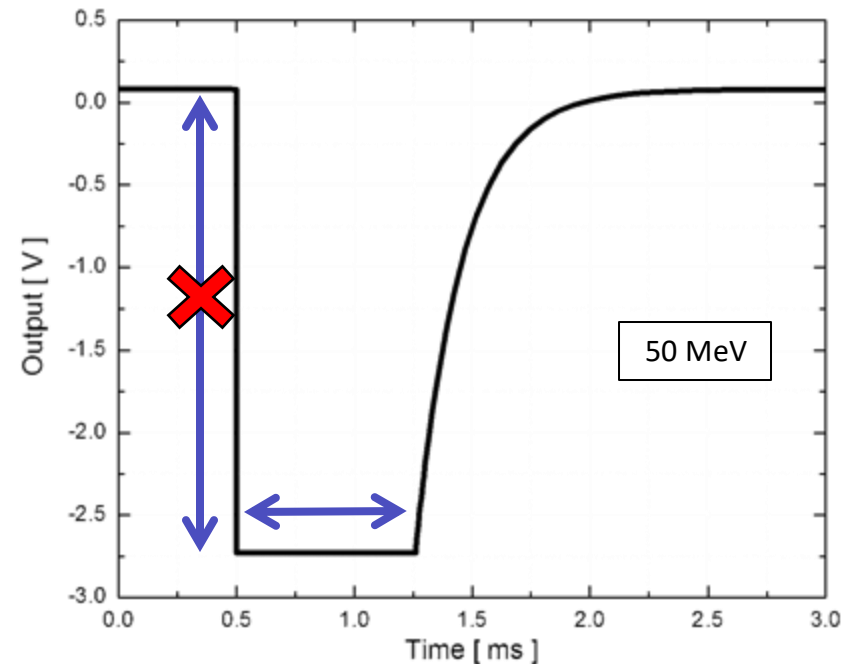


$$i(t) = Q \cdot \delta(t)$$

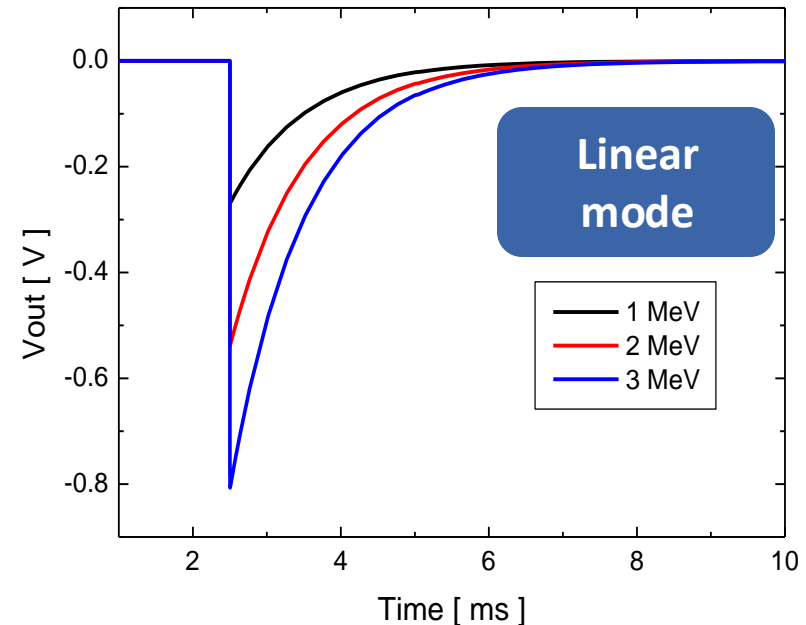
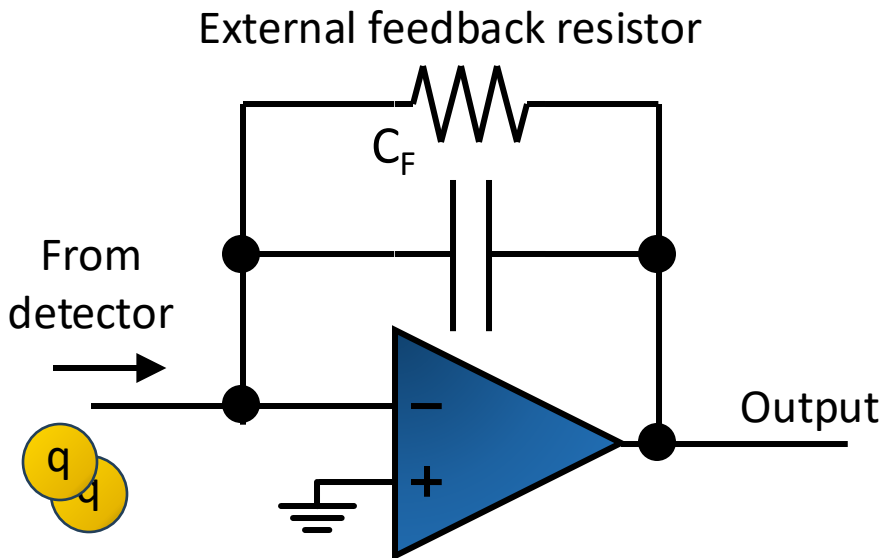
$$Q_{Emitted} \propto E_{Event}$$



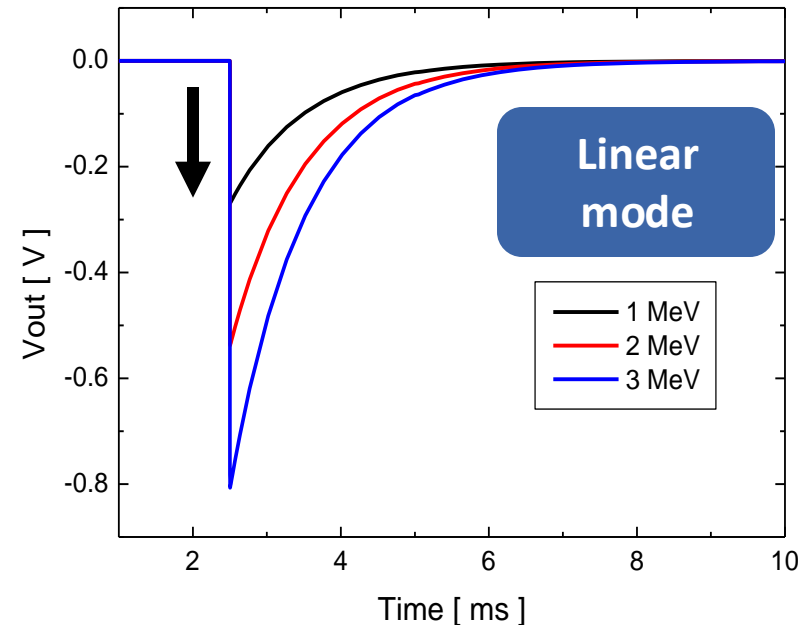
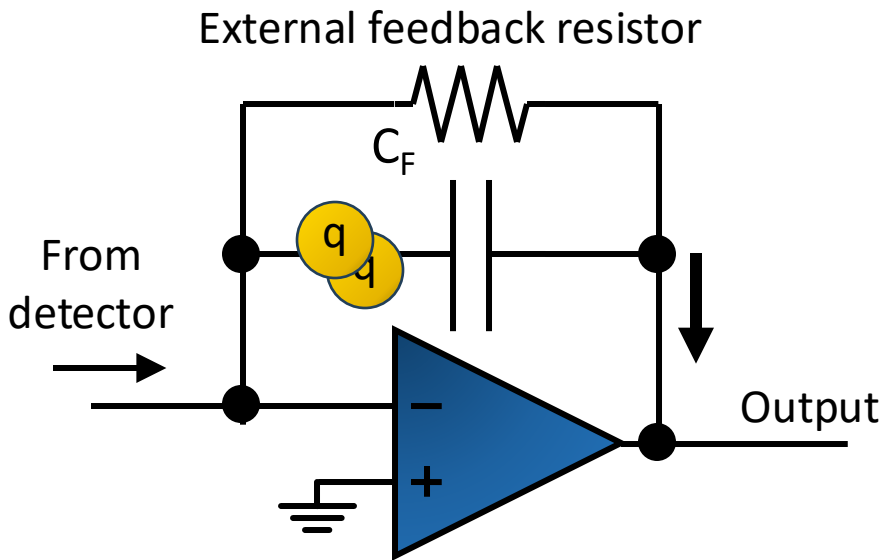
- Signal amplitude proportional to the energy deposited in the detector
- Decay constant determined by the RC product of the CSP



- Corrupted energy information
- Dead time that can be much longer than the decay constant of the preamplifier

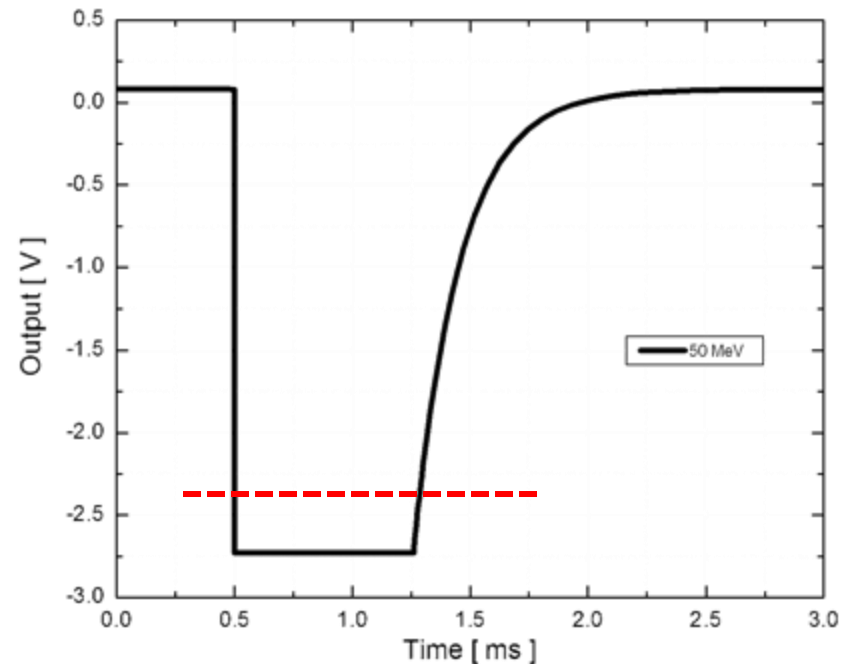
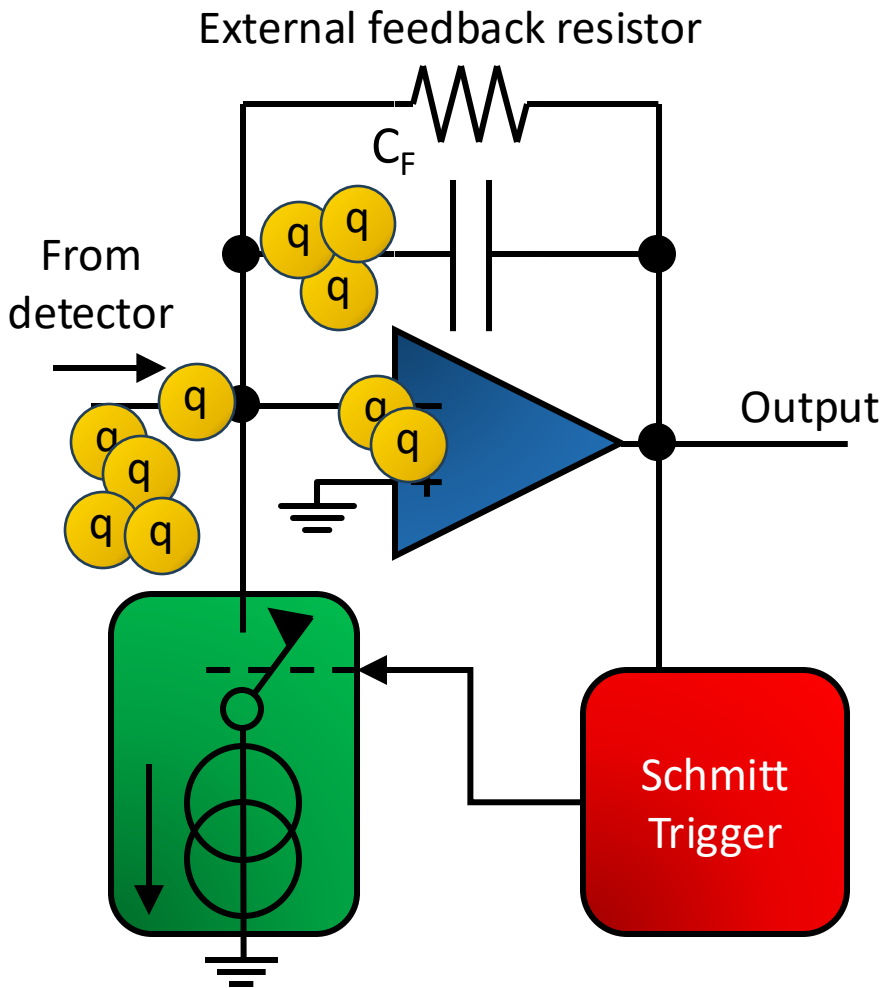


- The Fast-Reset preamplifier is a CSP equipped with a Schmitt Trigger and a Current sink
- For under-threshold signals it works like a normal CSP
- In case of saturation the current sink is activated

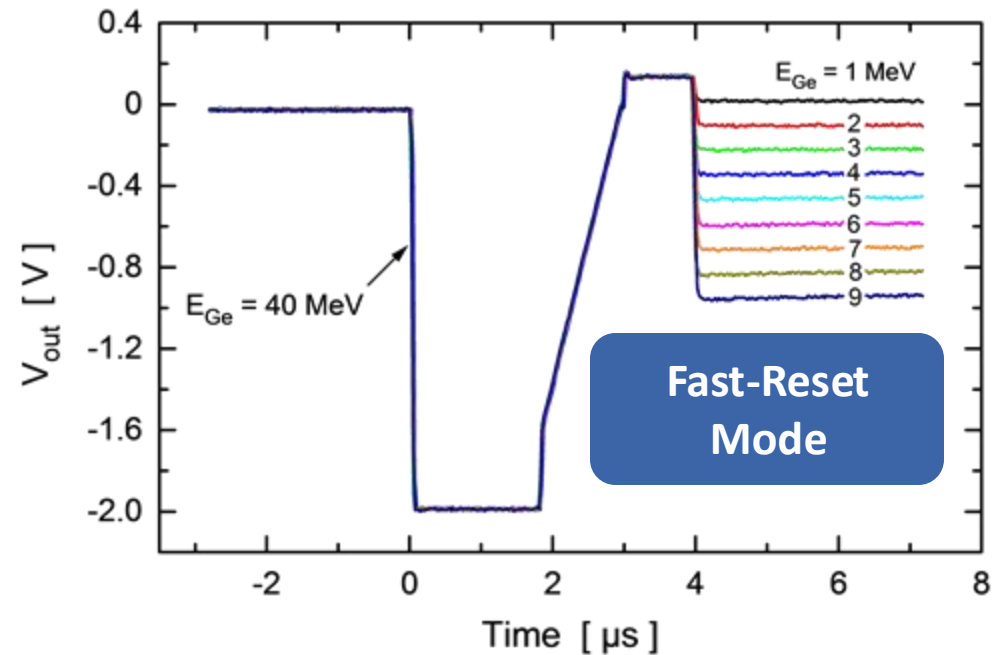
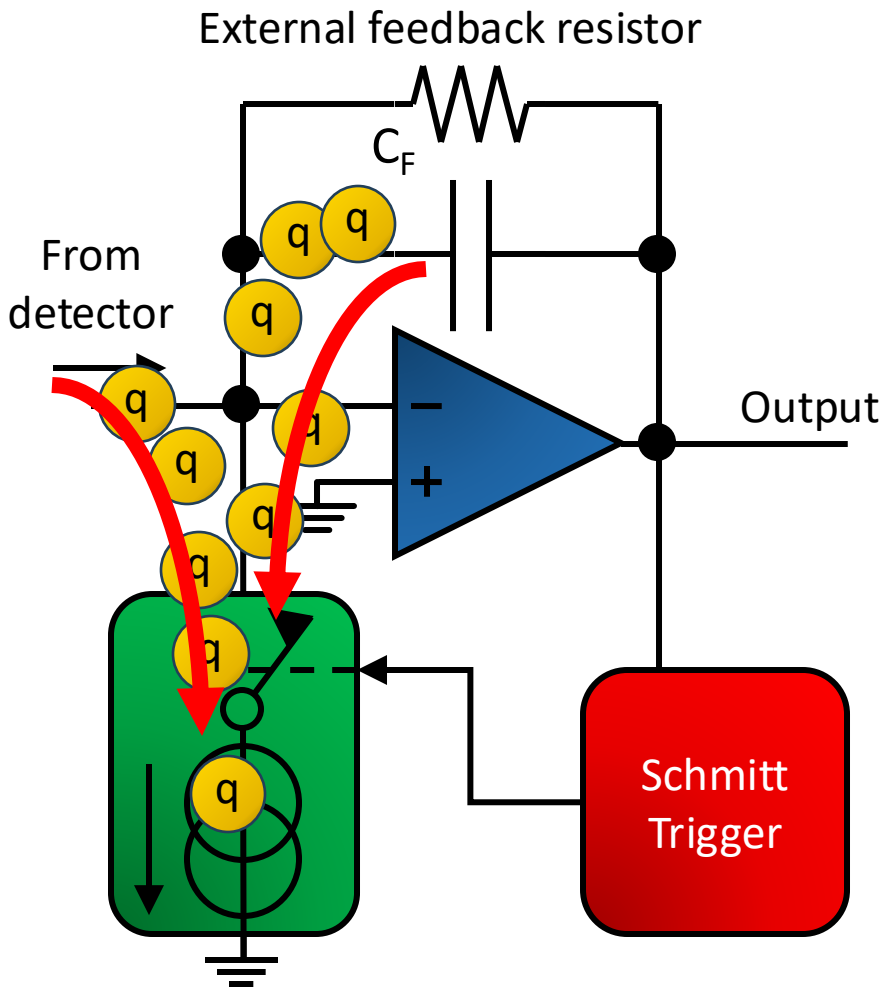


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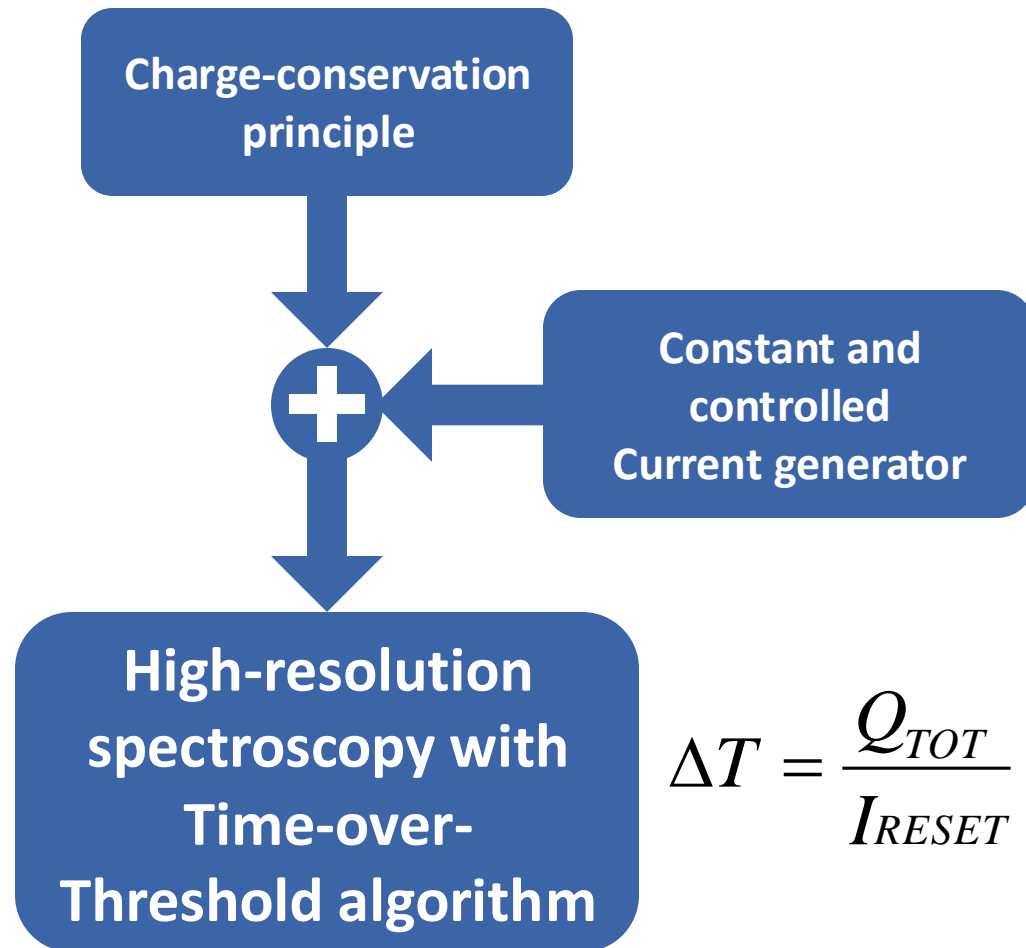
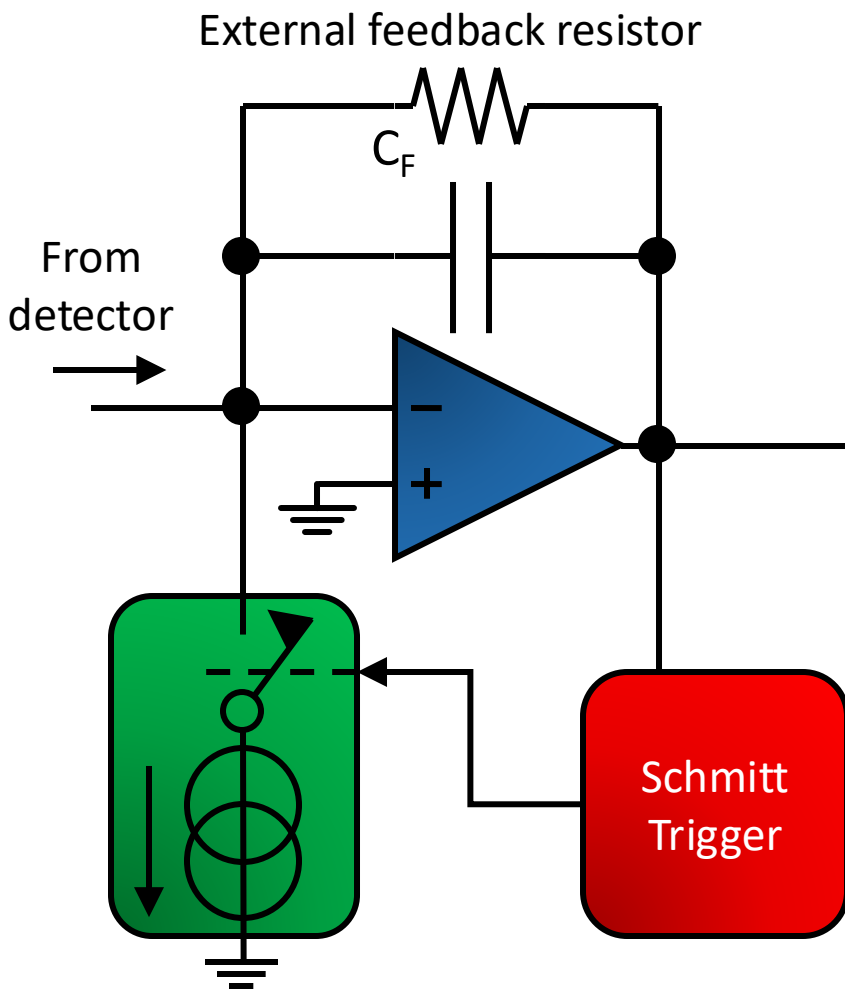




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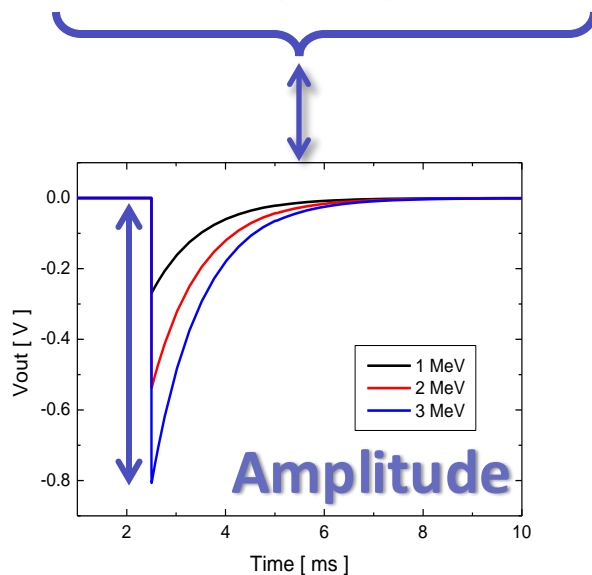
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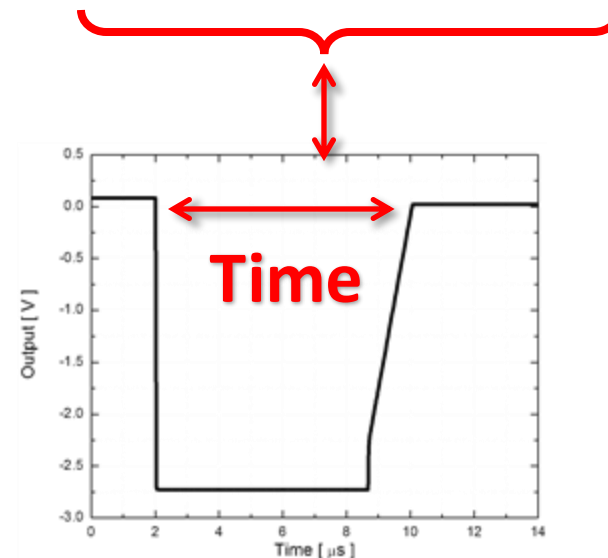


Combining offline the information collected with the two operating modes we can reconstruct the energy spectrum over an extended range

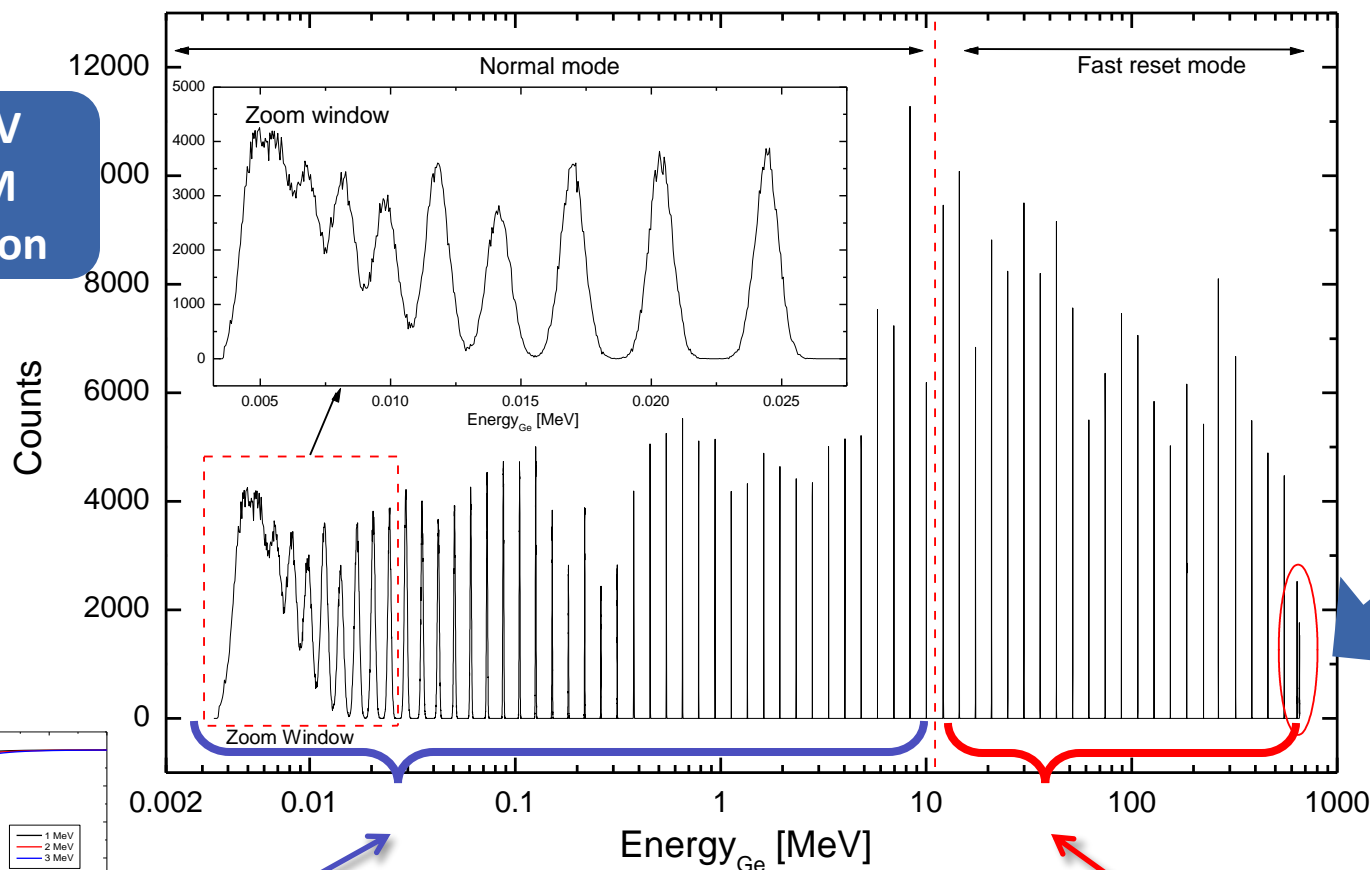
Energies under the  
saturation threshold  
 $< 10$  MeV



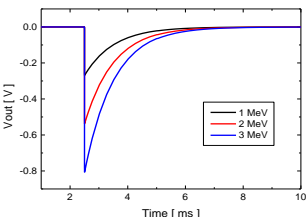
Energies over the  
saturation threshold  
 $> 10$  MeV



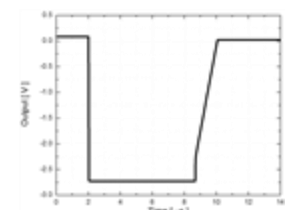
1.1 keV  
FWHM  
resolution



How can a saturated preamp saturate?

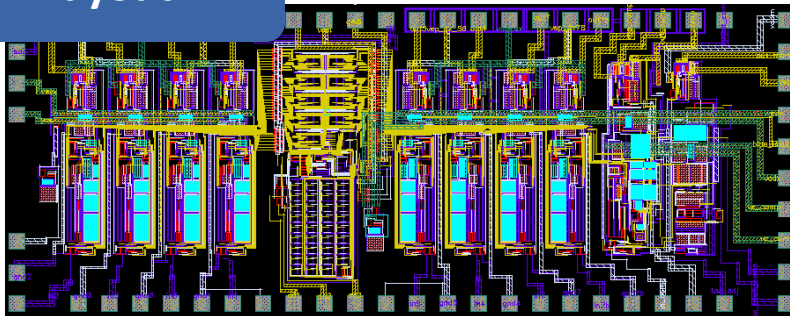


Amplitude

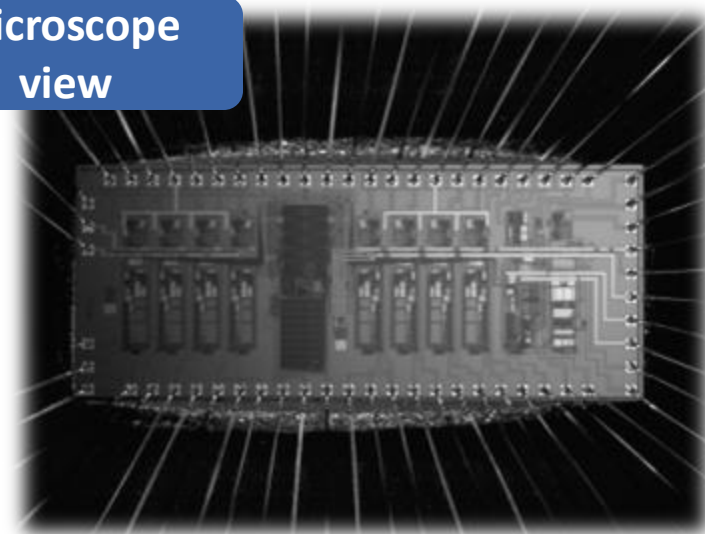


Time to Amplitude

Layout



Microscope  
view



- Realized in AMS C35 technology
- 8 channels for anodic signals and 1 channel for cathodic ones
- Power consumption: 12 mW/ch
- Risetime: 10 ns (4 pF det. And 1 pF FB)
- Power supply:  $\pm 2.5V$
- Area = 10mm<sup>2</sup>

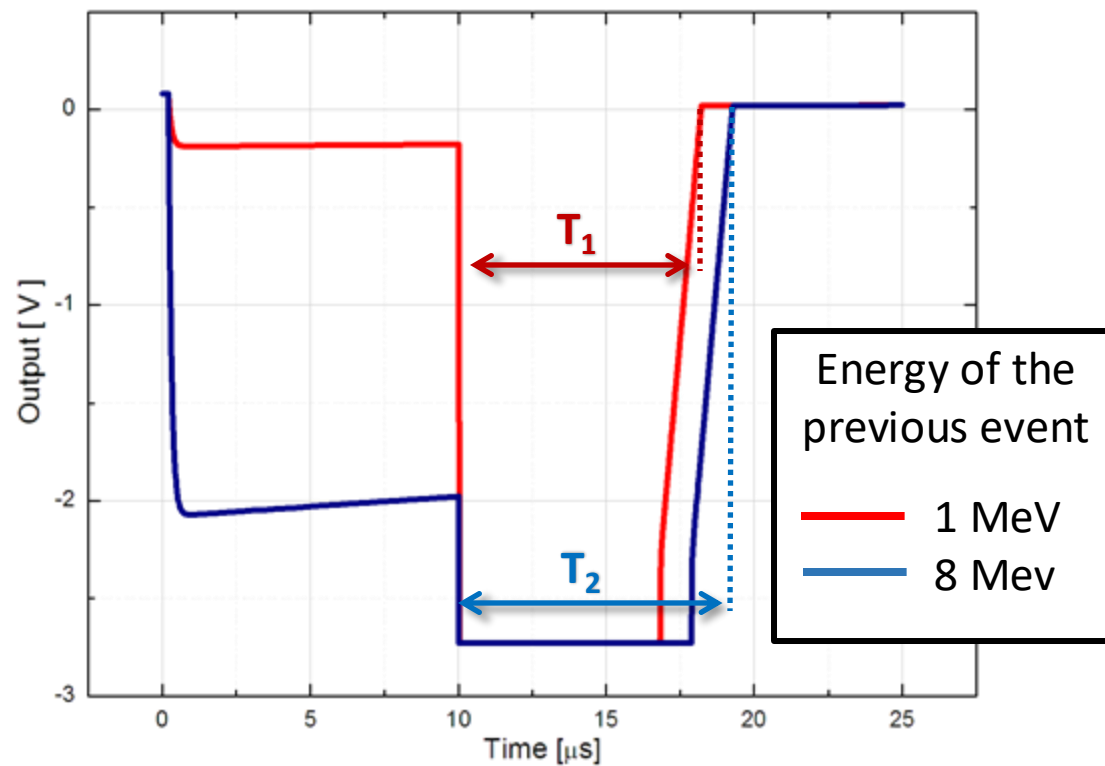
- Carrier: PLCC68 or bare die
- Digital slow control with I2C engine
- Separate power rails for cross-talk reduction
- Equipped with Fast-Reset circuit
- Only one external component: the feedback resistor

# Active Stoppers for Decay Experiments

## Dependency of the energy measurement from the baseline value



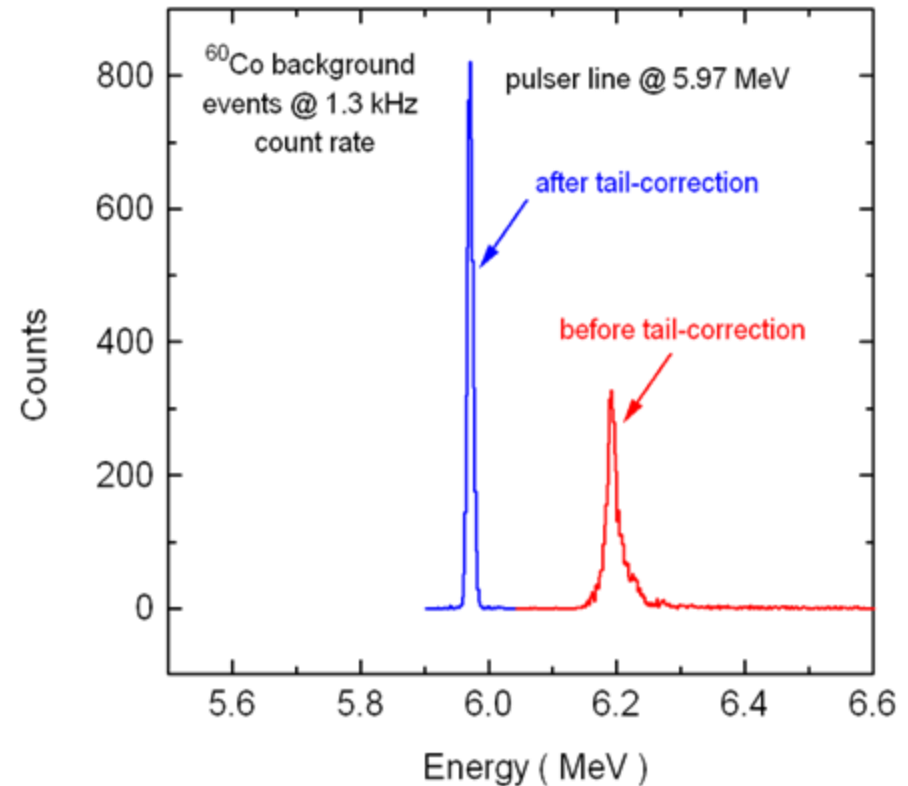
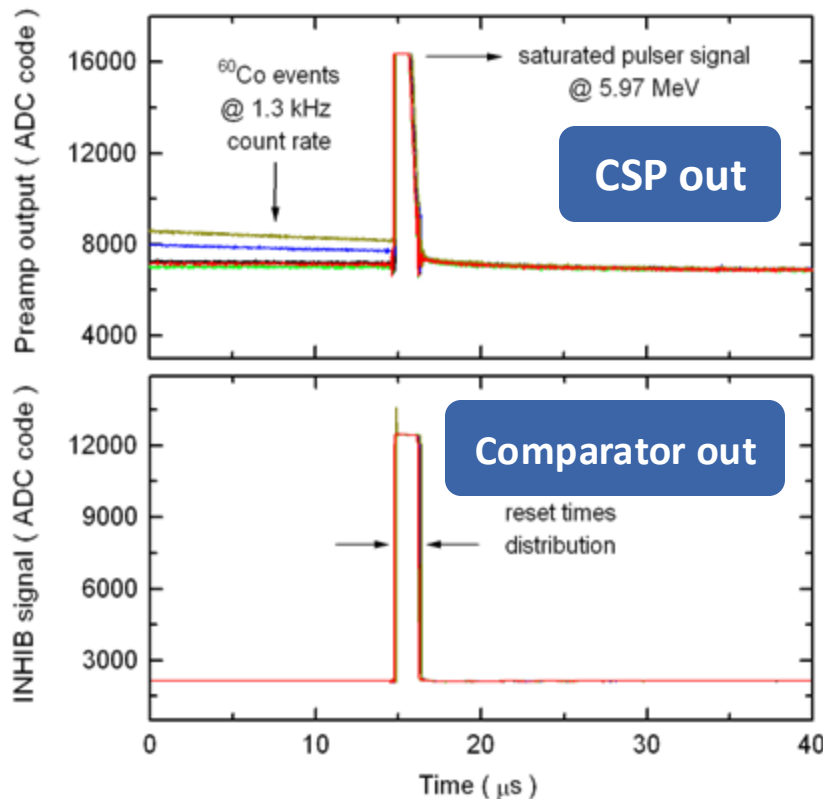
- The digital TOT signal depends on the residual charge on  $C_F$  before the reset process
- **Need for an algorithm to correct this dependency**
- Off-line digital correction: easy to implement but expensive in computational terms





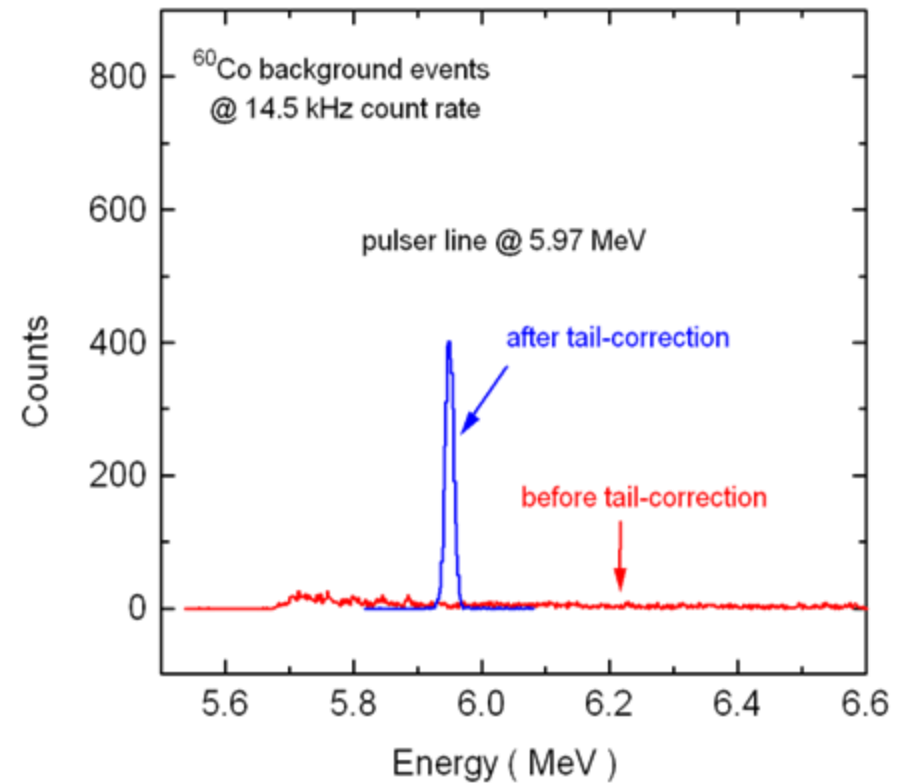
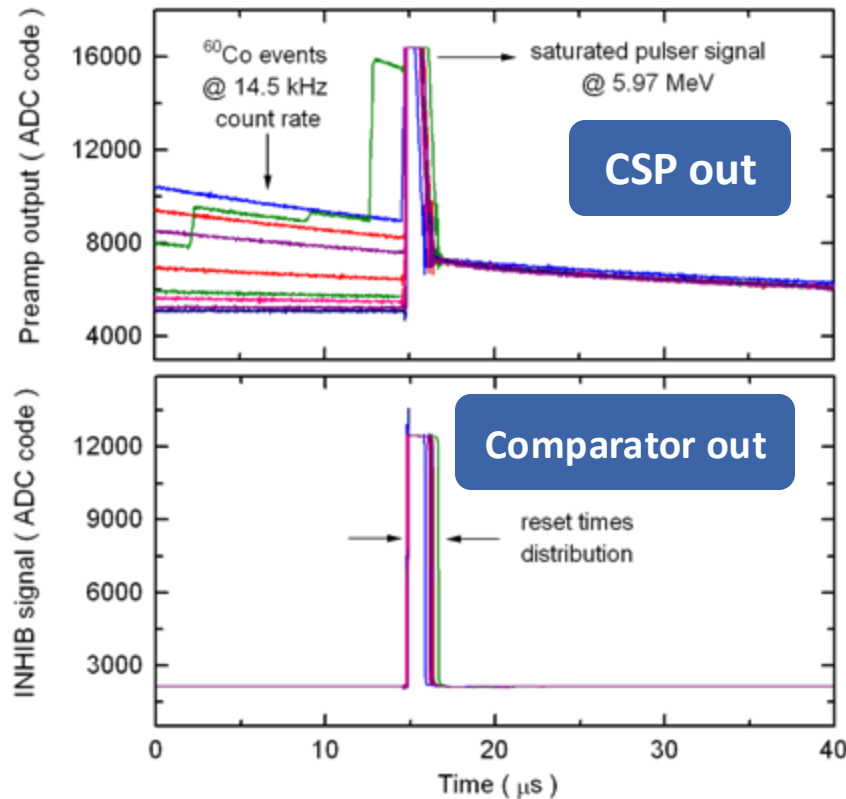
## Pile-up effects at medium counting rate

1.3 KHz



## Pile-up effects at high counting rate

14.5 KHz



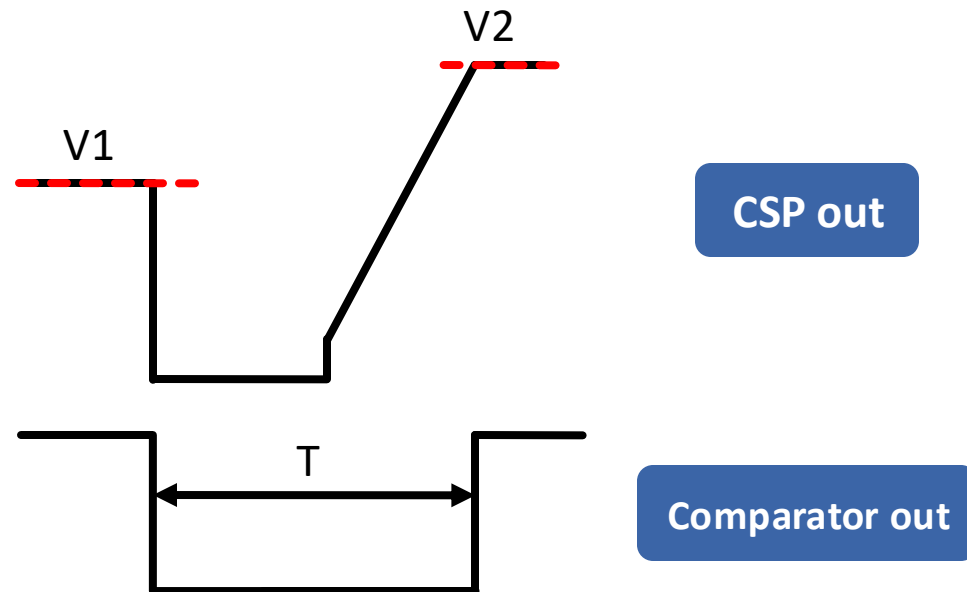
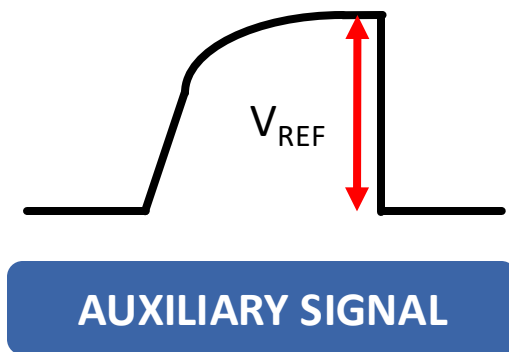


# An algorithm to correct the spectra from the baseline dependency



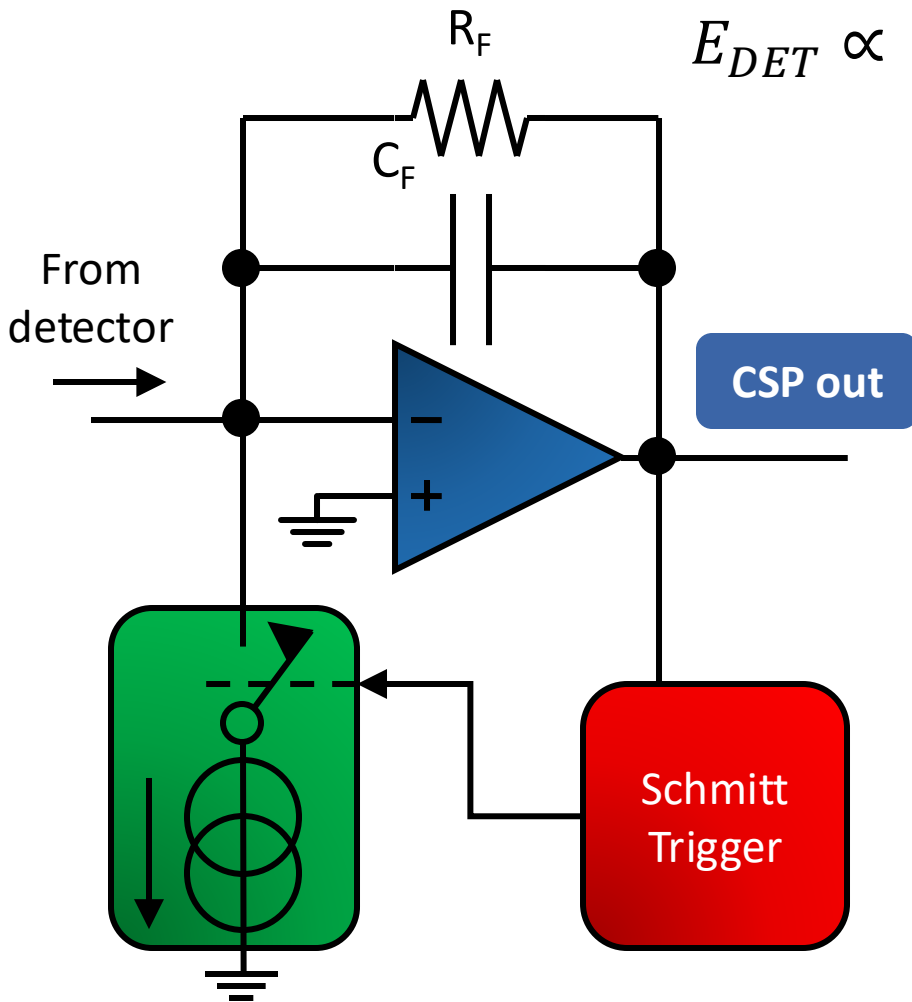
$$E_{DET} \propto V_{REF} = \alpha [I_{RESET} \cdot T - C_F \cdot (V_2 - V_1)]$$

Need to generate an auxiliary signal  
with amplitude  $V_{REF}$  directly  
proportional to the energy of the last  
physical event (and that doesn't depend  
on the residual charge of past events!)

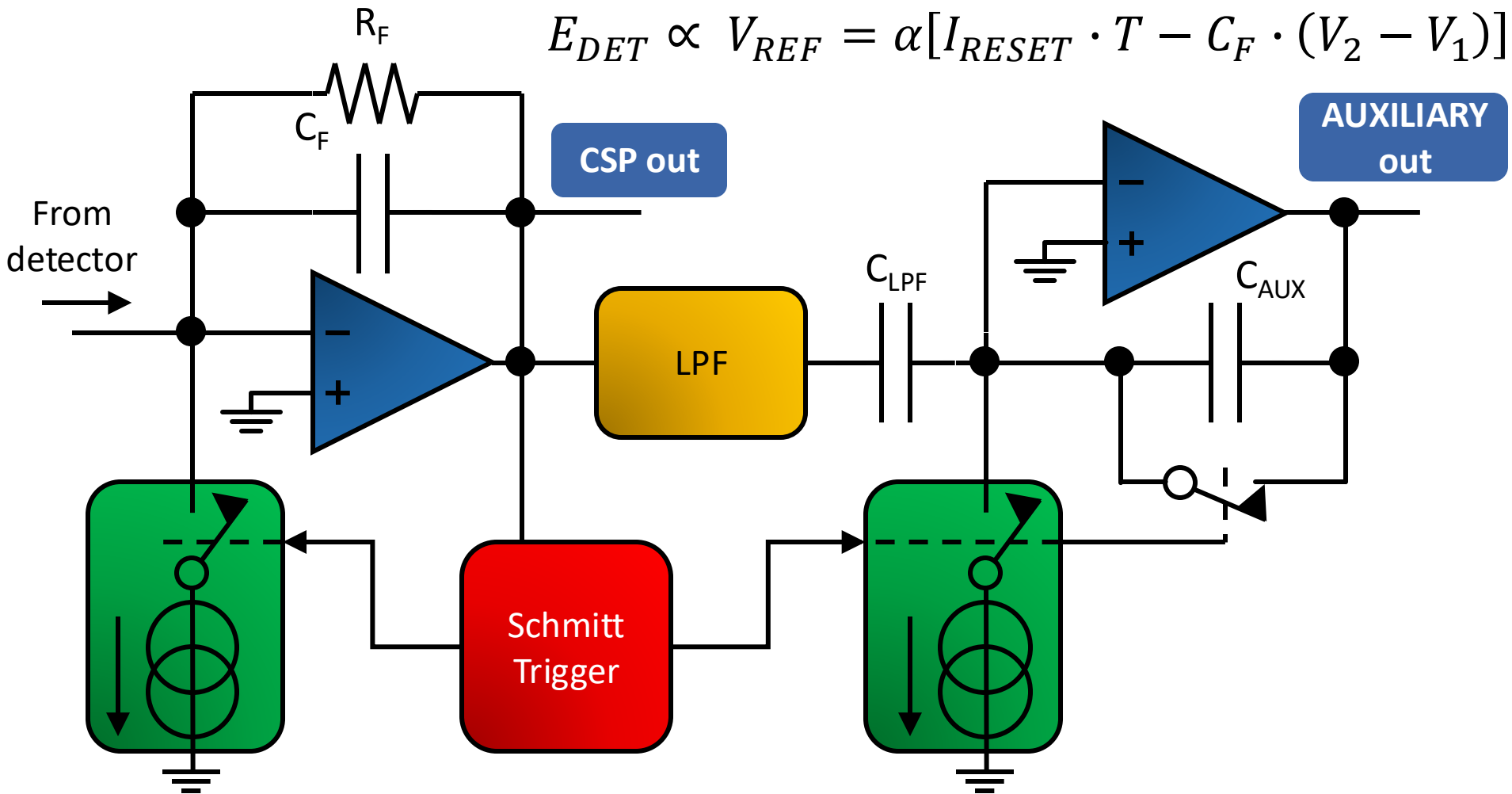


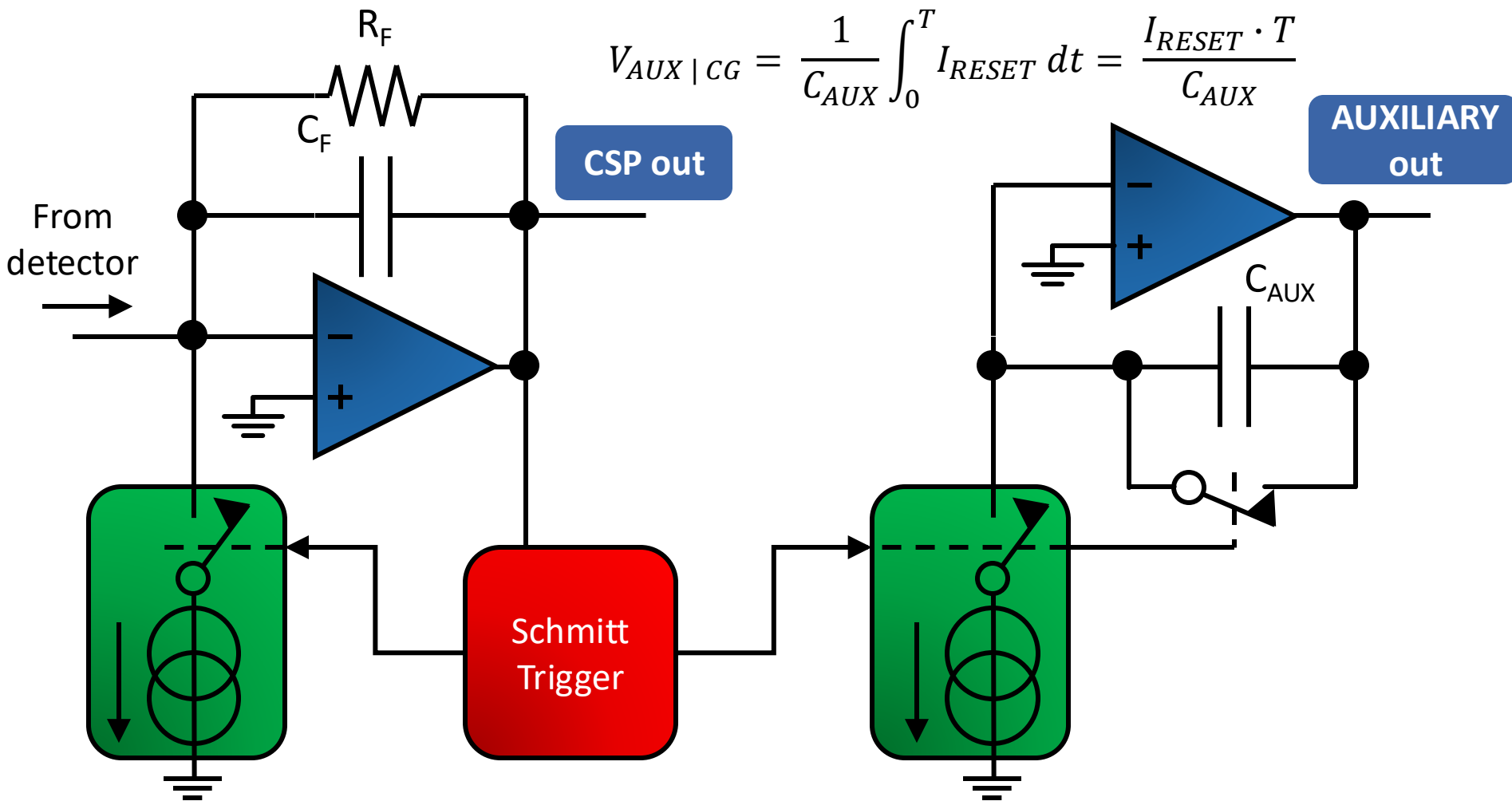


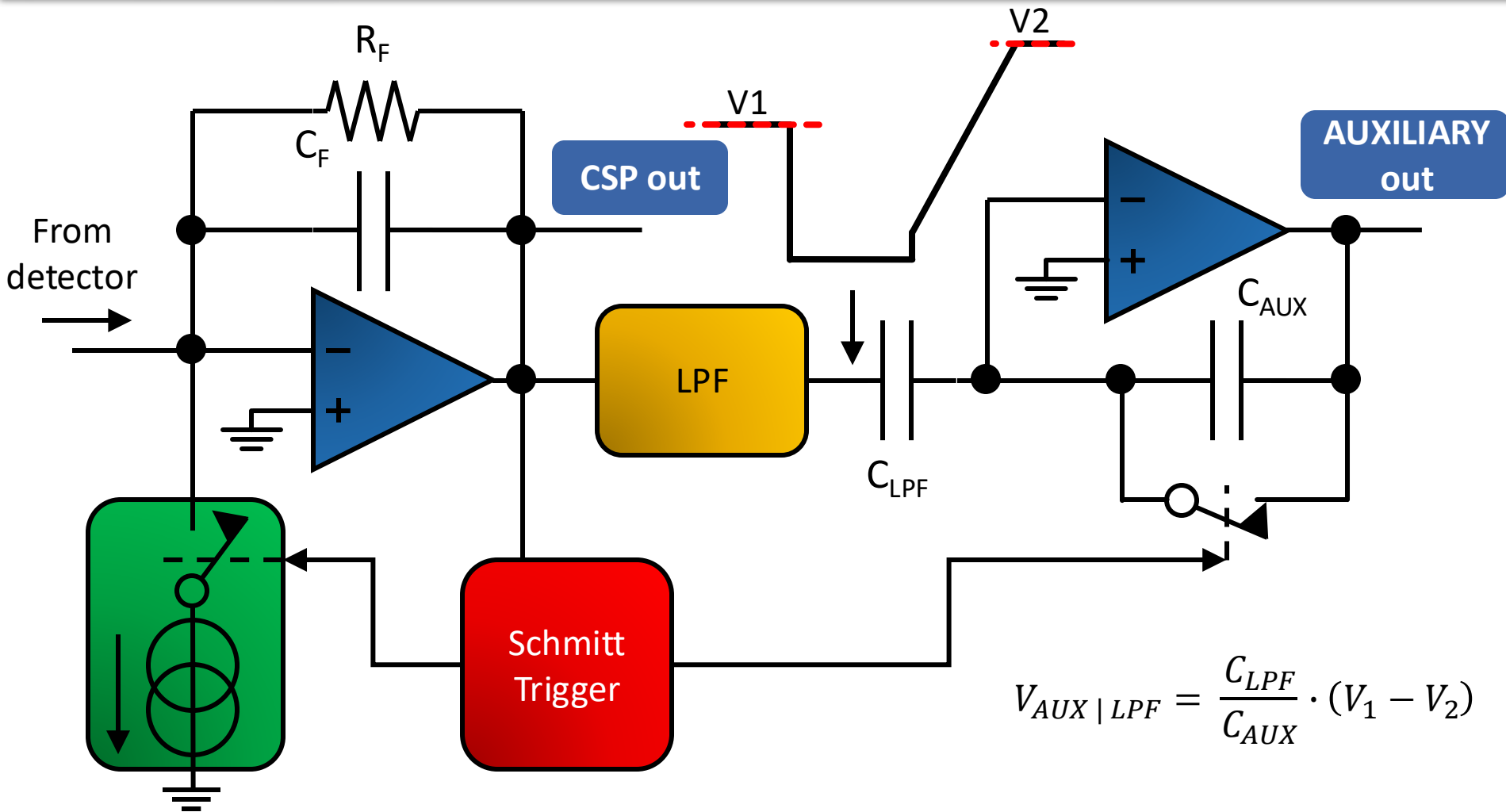
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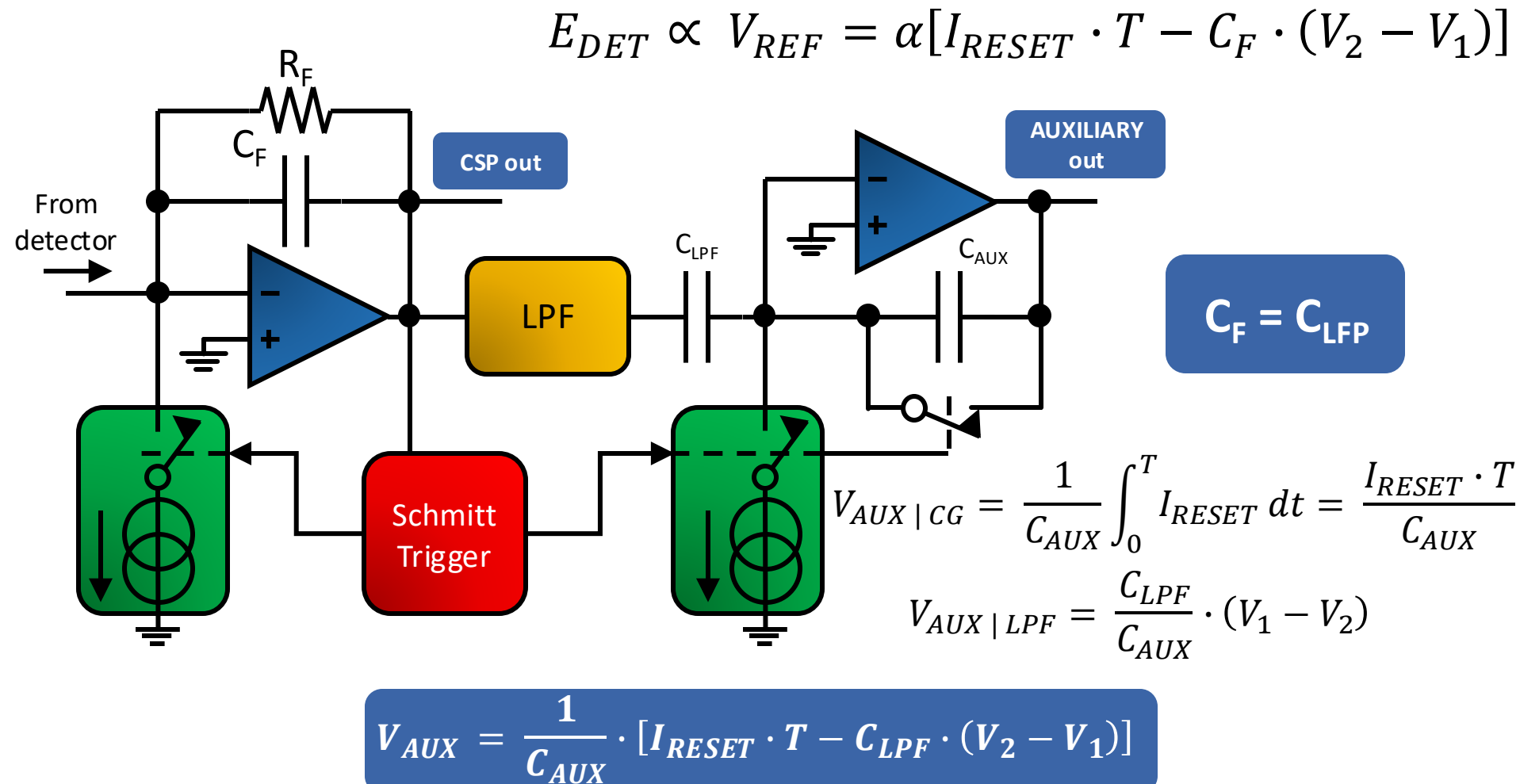


$$E_{DET} \propto V_{REF} = \alpha [I_{RESET} \cdot T - C_F \cdot (V_2 - V_1)]$$

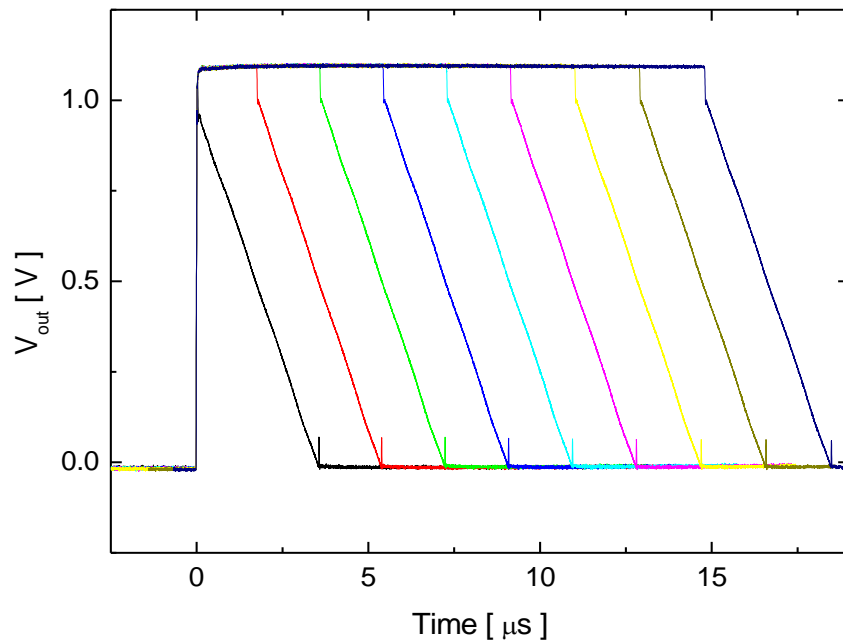






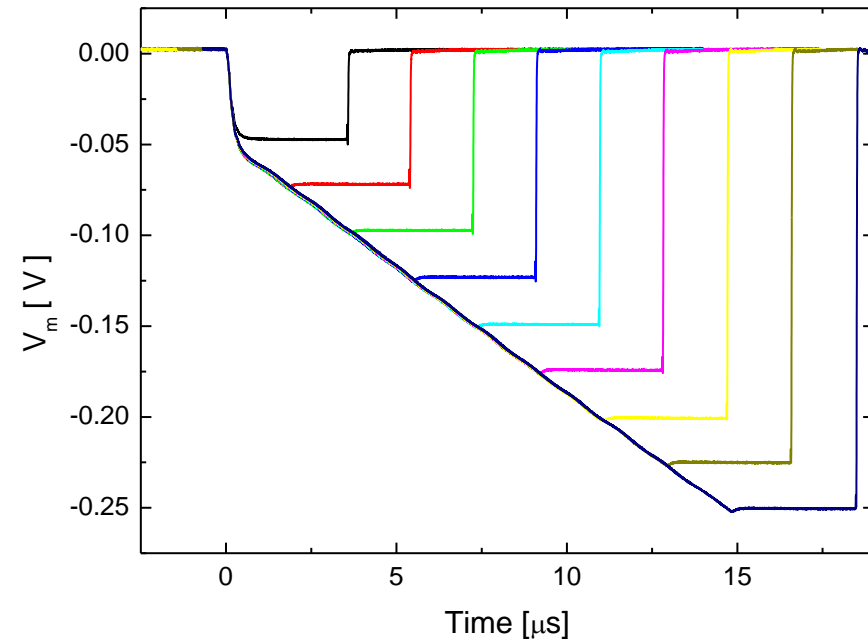


### CSP out



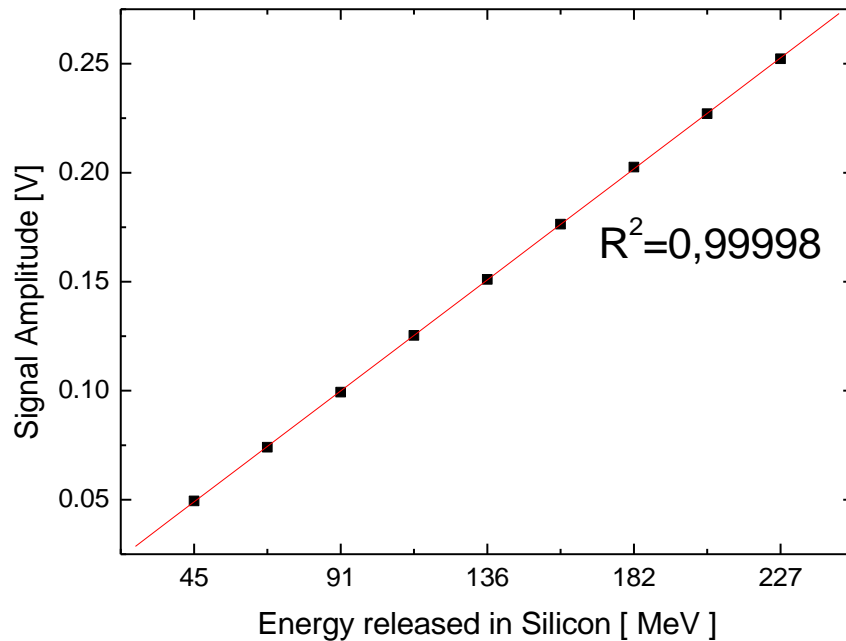
- 2 pC to 10 pC charge signals injected on the input node of the CSP with a pulser through a 1 pF test capacitor

### AUXILIARY out



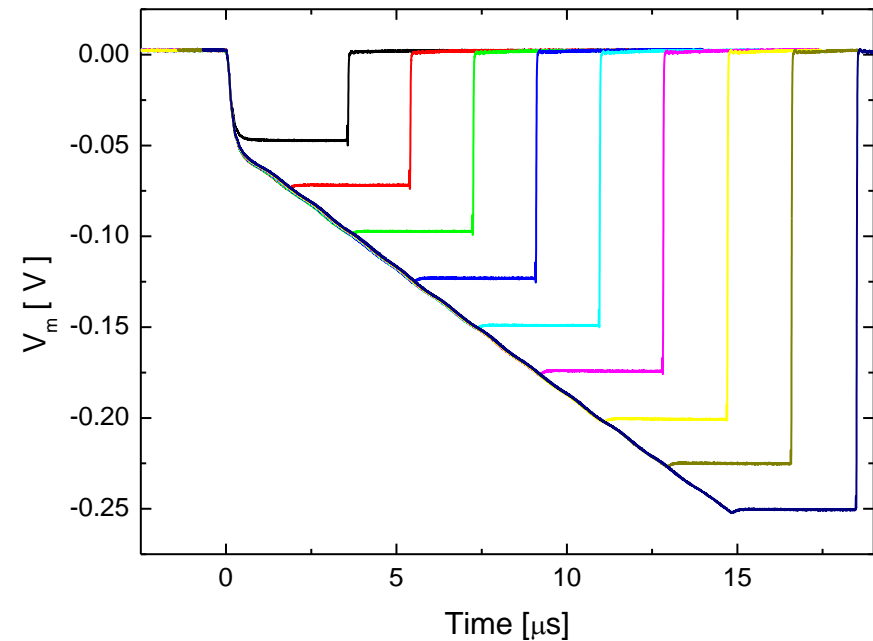
- The auxiliary TAC structure produces signals that are linear in amplitude with the energy!

Linear fit



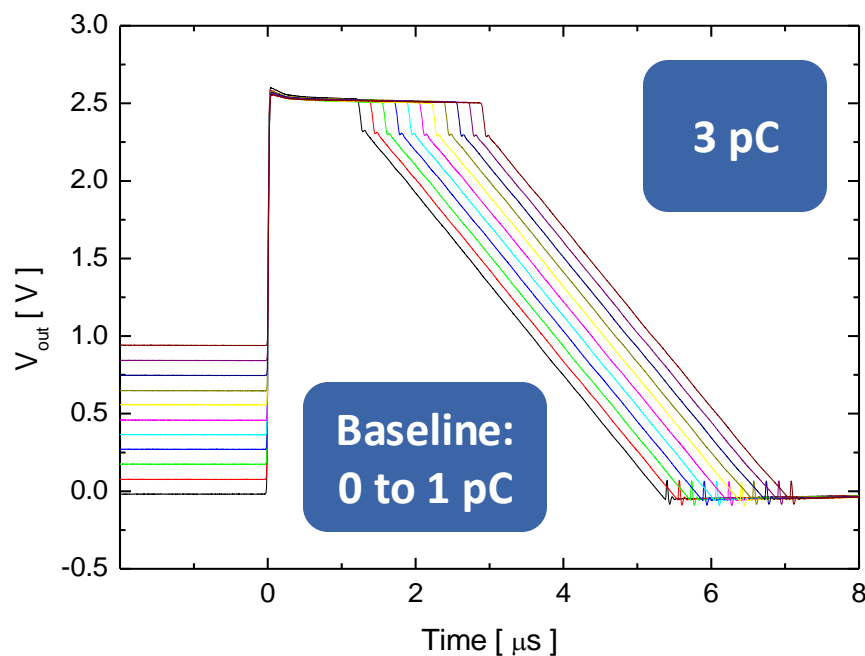
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AUXILIARY  
out

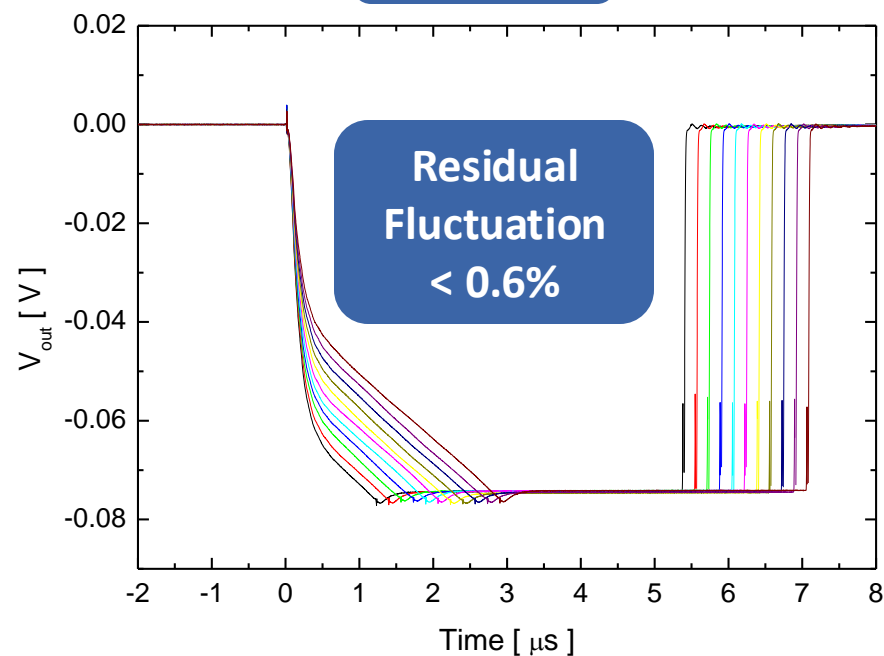


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### CSP out



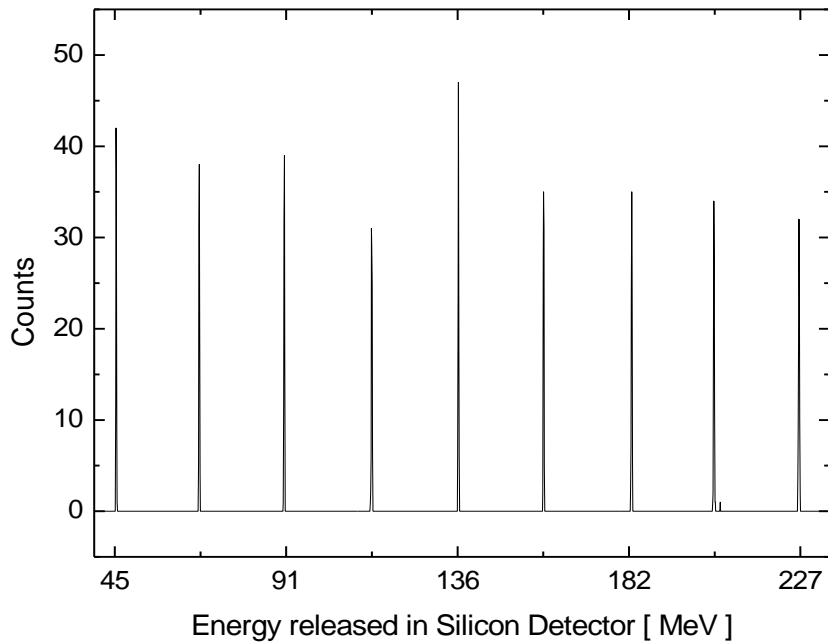
### AUXILIARY out



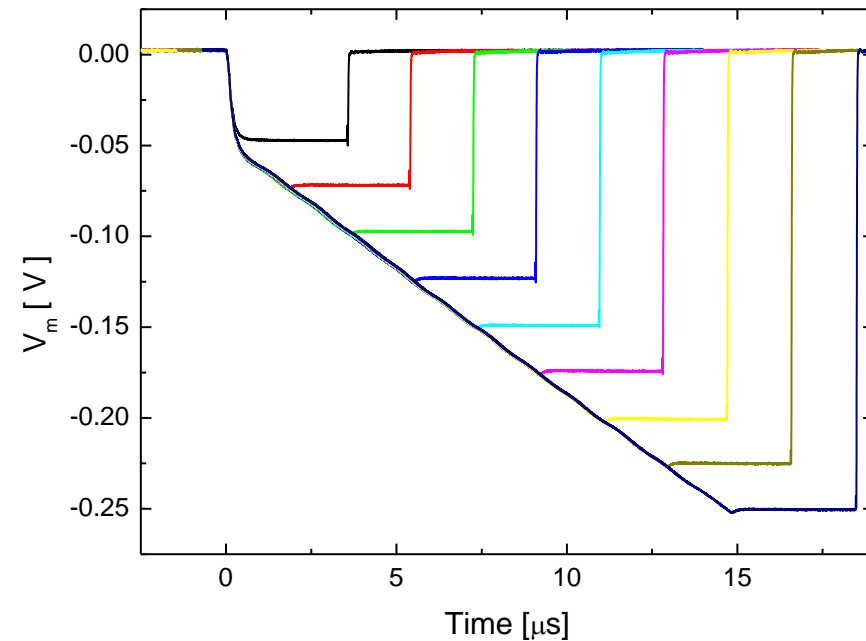
- 3 pC charge signals injected on the input node of the CSP with a pulser through a 1 pF test capacitor. From 0 to 1pC of residual charge on the input node

- The auxiliary TAC structure produces signals that change in shape but keep the same amplitude!

### Amplitude histogram

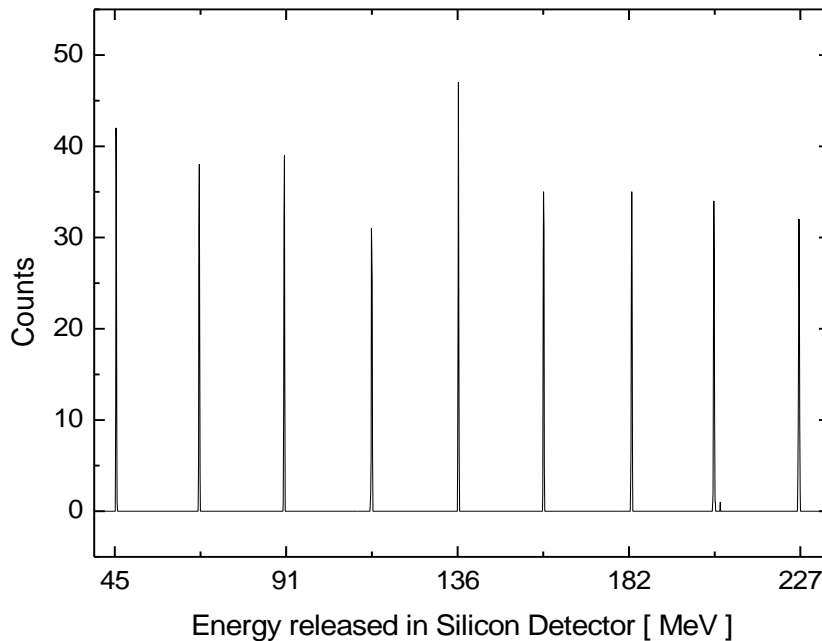


### AUXILIARY out



- 2 pC to 10 pC charge signals injected on the input node of the CSP with a pulser through a 1 pF test capacitor. 100 signals acquired for each peak

Amplitude histogram

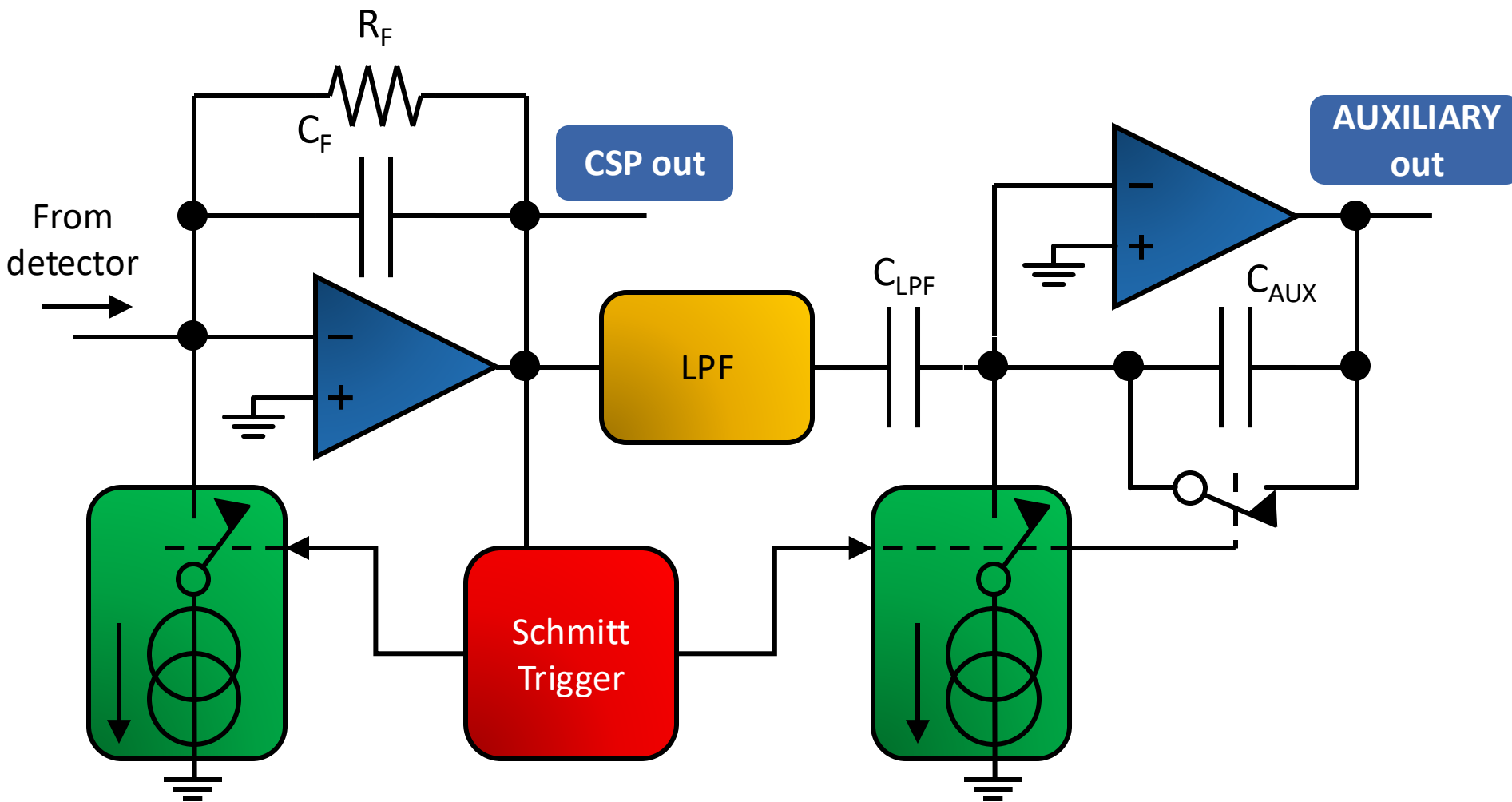


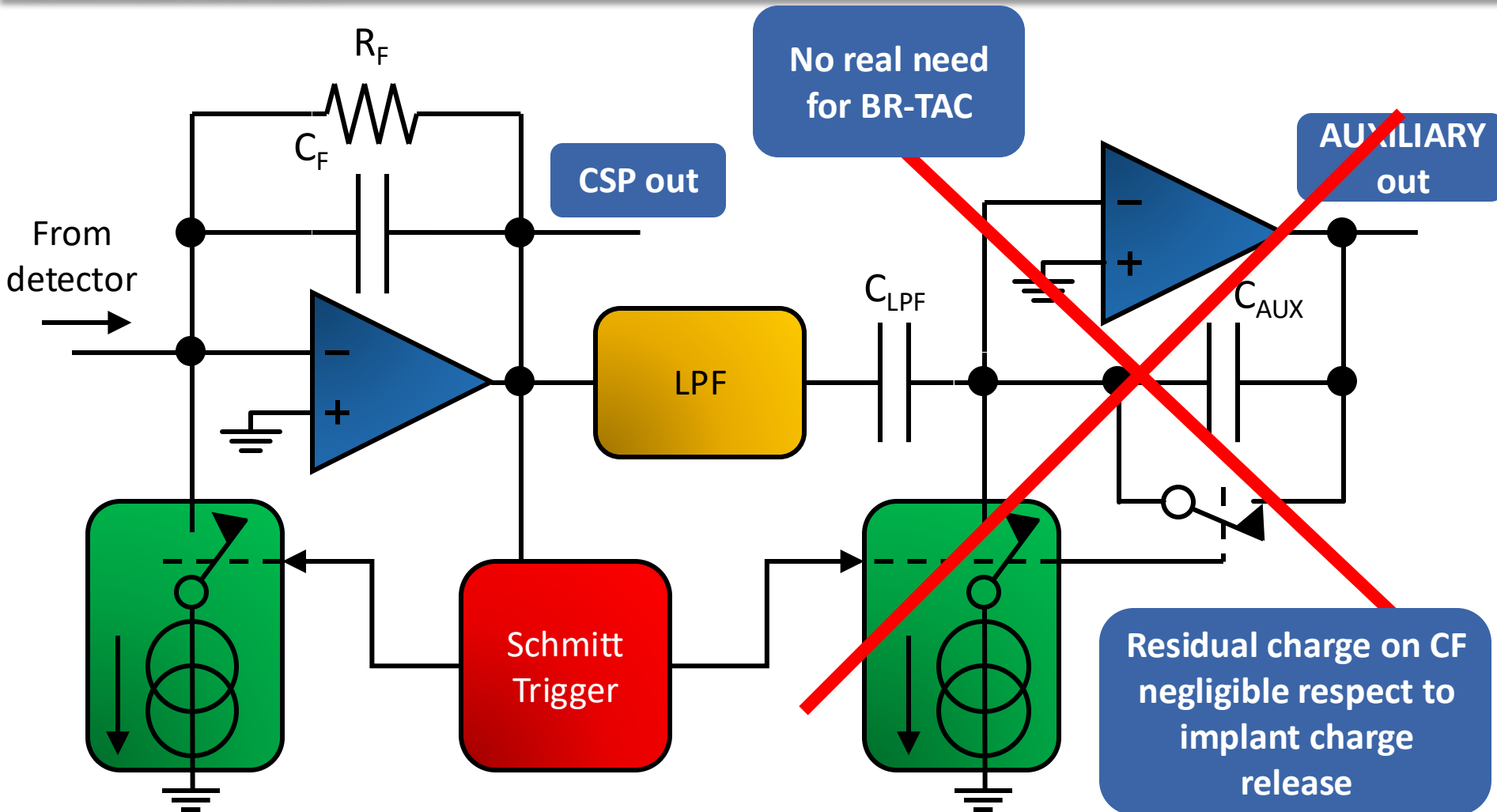
Peak resolutions

Input Charge [pC]	Signal amplitude [mV]	Peak width FWHM [mV]	Peak width FWHM [%]
2	49.394	2.10E-01	0.43
3	74.073	2.43E-01	0.33
4	99.329	2.29E-01	0.23
5	125.33	2.96E-01	0.24
6	151.08	2.47E-01	0.16
7	176.39	2.27E-01	0.13
8	202.56	2.69E-01	0.13
9	227.03	5.35E-01	0.24
10	252.27	3.47E-01	0.14

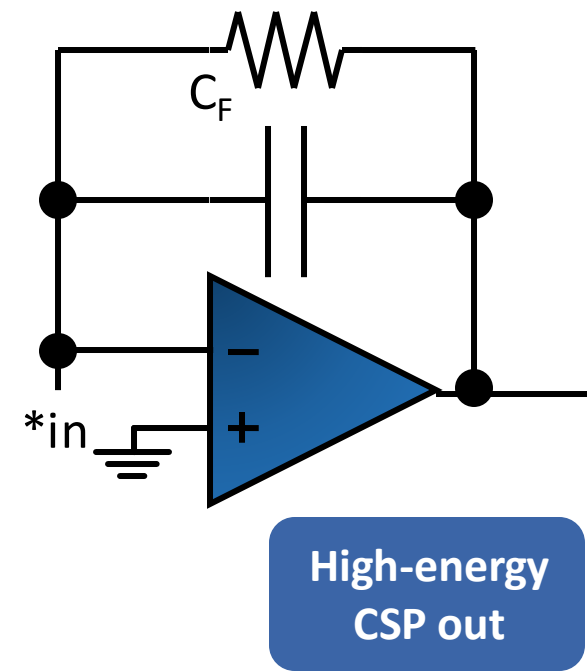
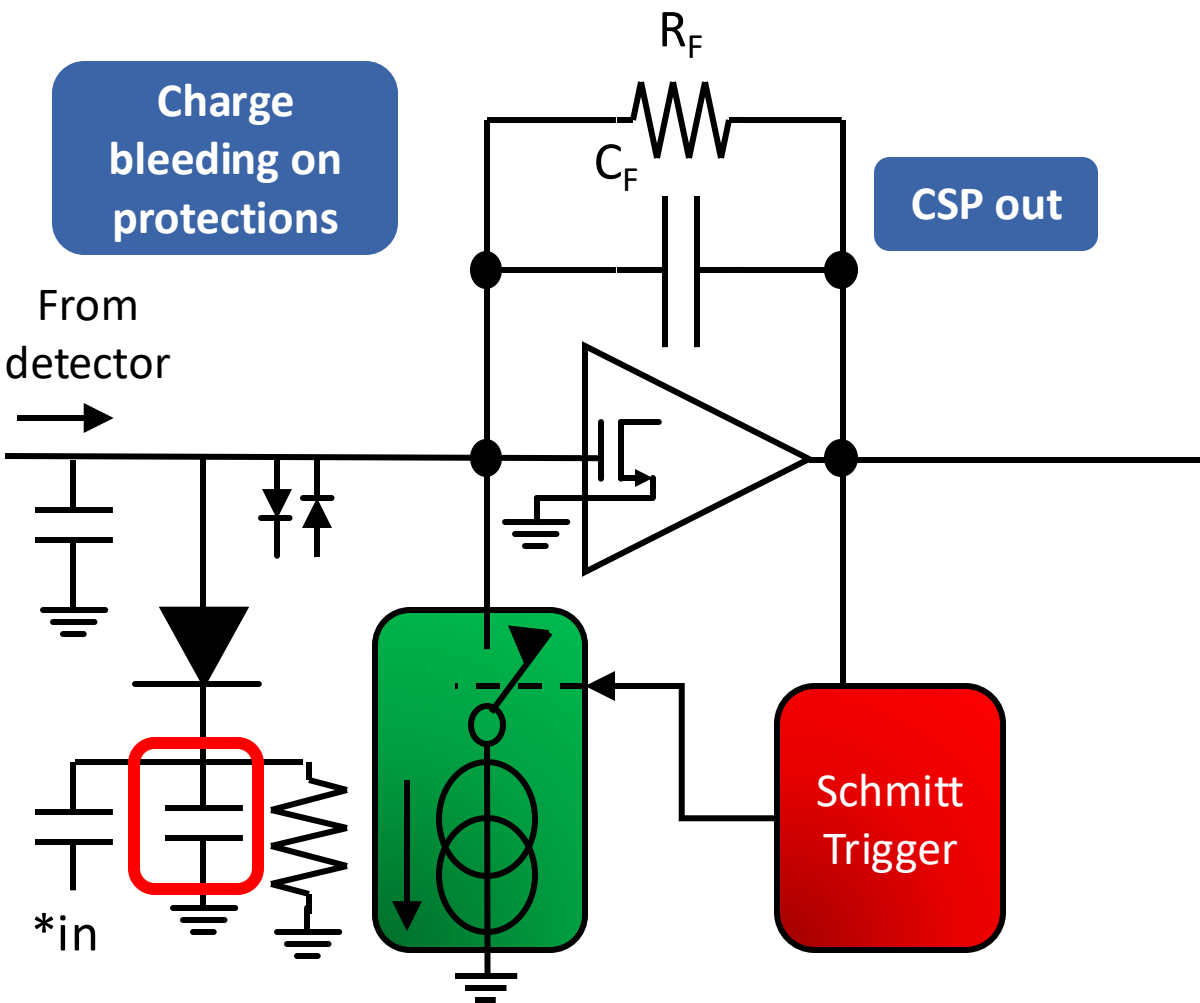
**BEST-CASE RESOLUTION OF 0.13% FWHM OF THE TOTAL ENERGY!**

- 2 pC to 10 pC charge signals injected on the input node of the CSP with a pulser through a 1 pF test capacitor. 100 signals acquired for each peak

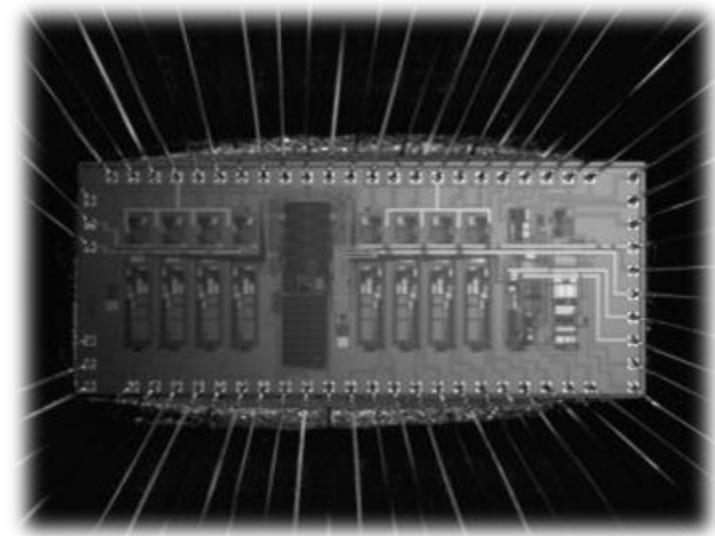
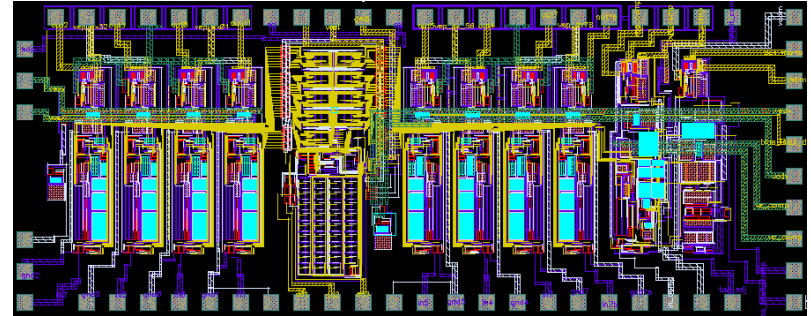








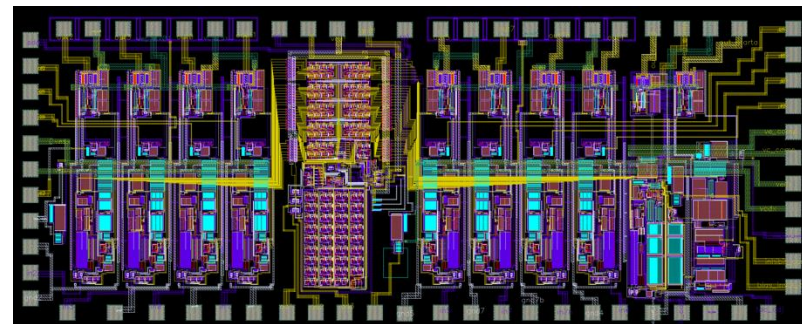
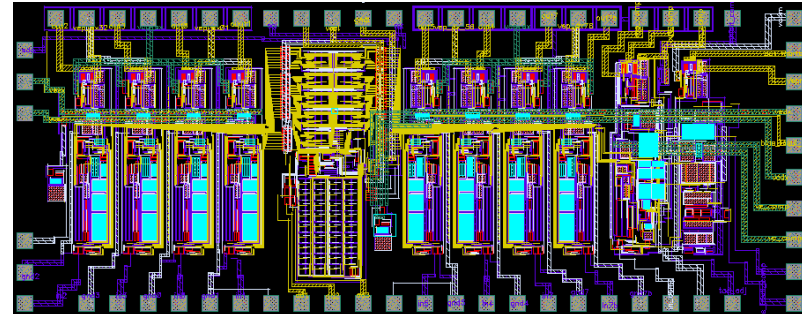
- A low-noise low-power CSP ASIC was presented with an innovative range-booster circuit
- The CSP meets the requirements of gamma spectroscopy (but also suitable for particle spectroscopy)
- The fast risetime enables to process the signals from this preamplifier with pulse-shape analysis algorithms
- An innovative technique was presented that extends the natural dynamic range of the preamplifier from 40 MeV to several hundreds of MeV
- The algorithm was implemented in an analog circuits that performs the operation on-line and is not influenced by the signal's baseline



New ASIC preamplifier for  
signals with opposite polarity:  
Ready, tested.

CSP implementation for decay  
spectroscopy applications

Experimental results: hopefully  
before end of 2025



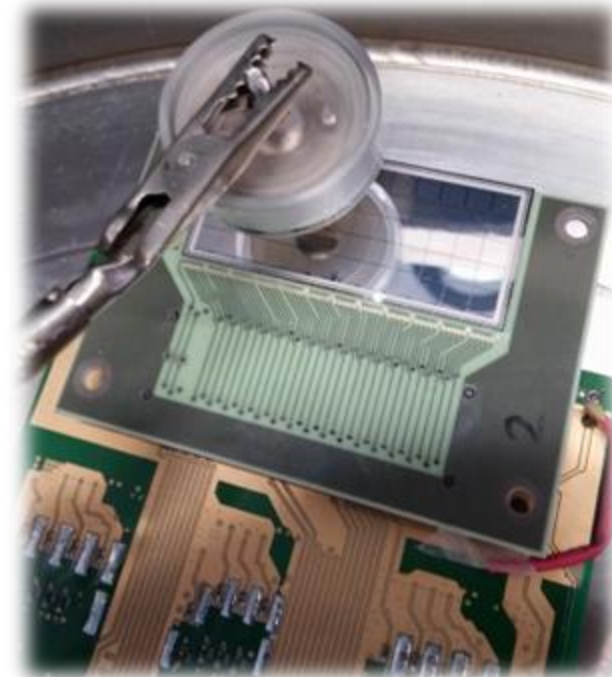
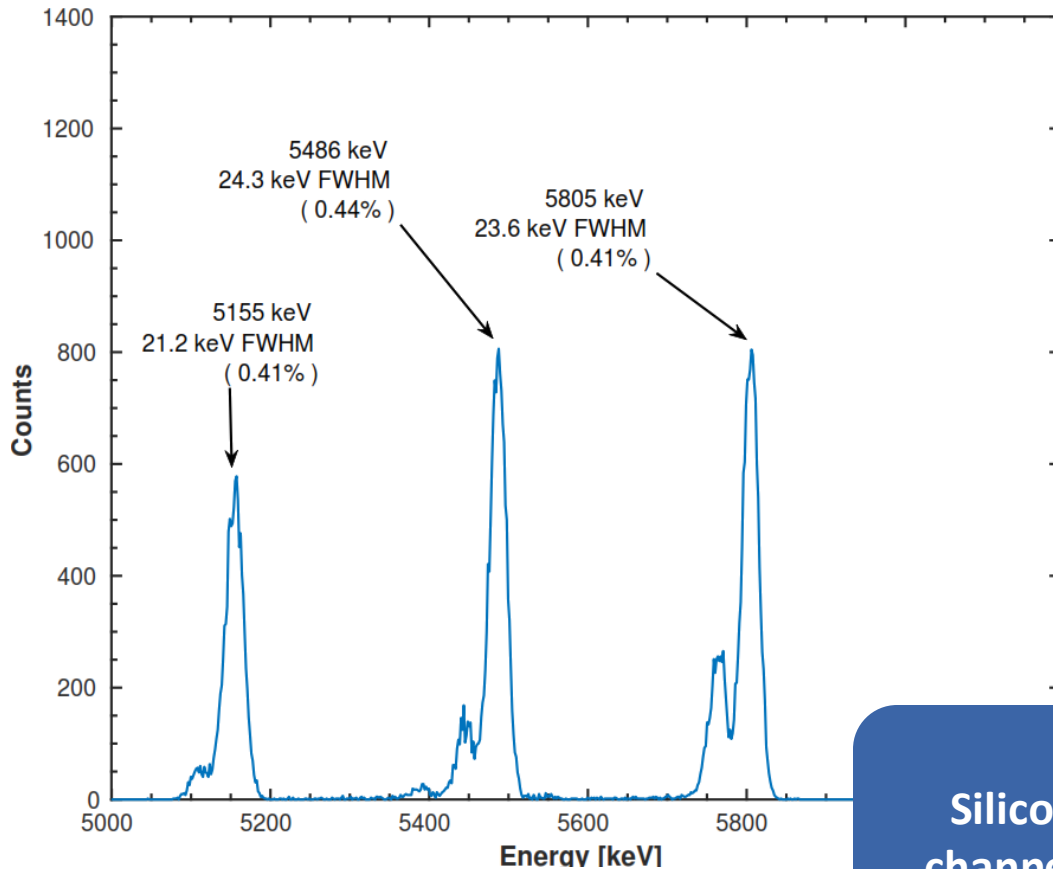
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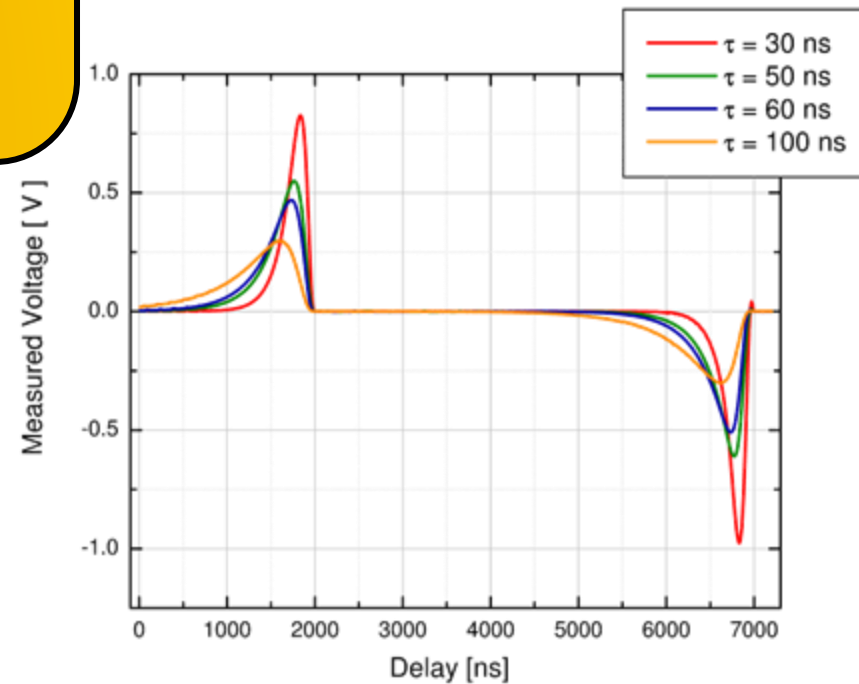
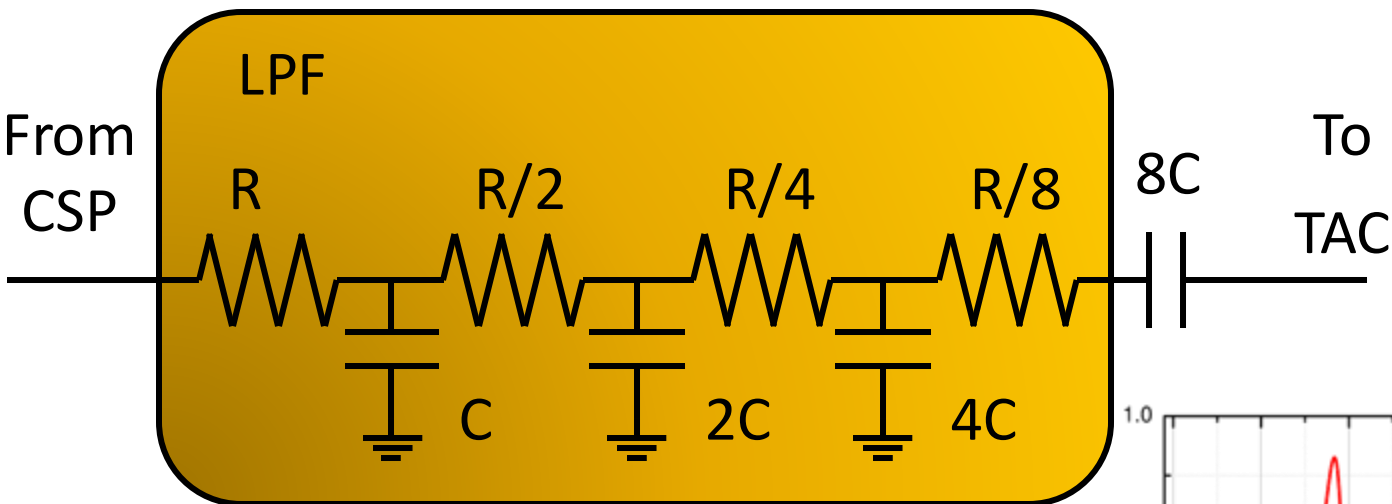
Thank you

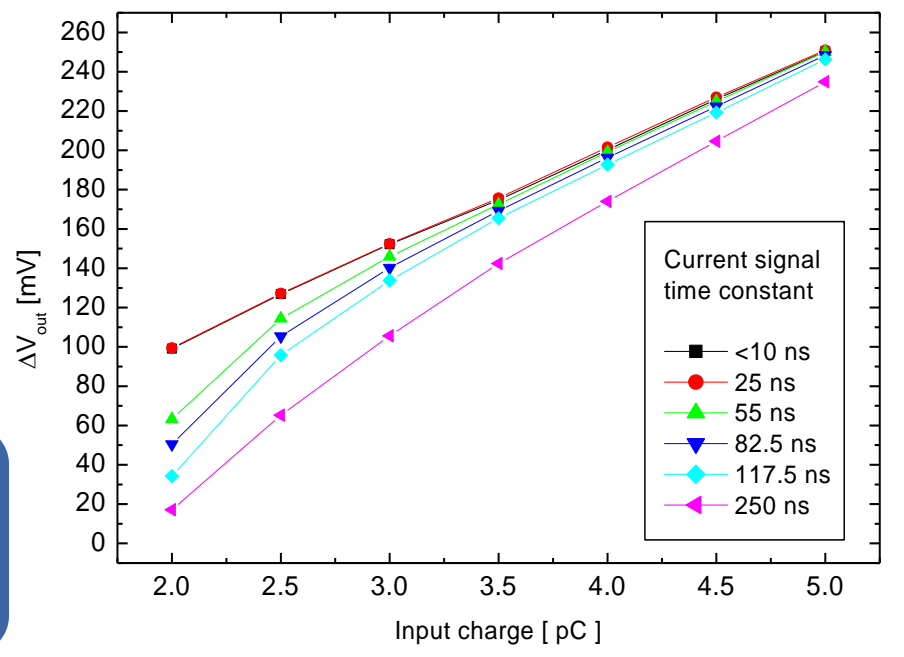
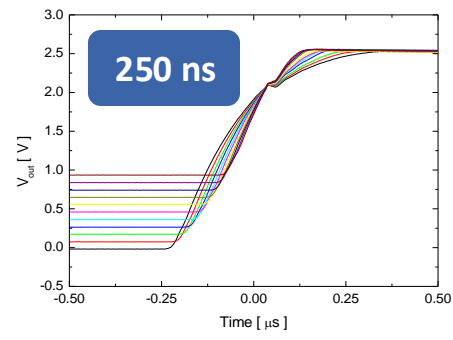
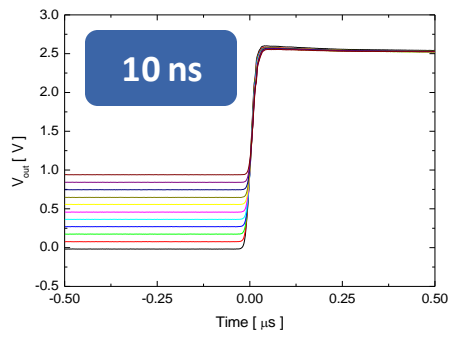
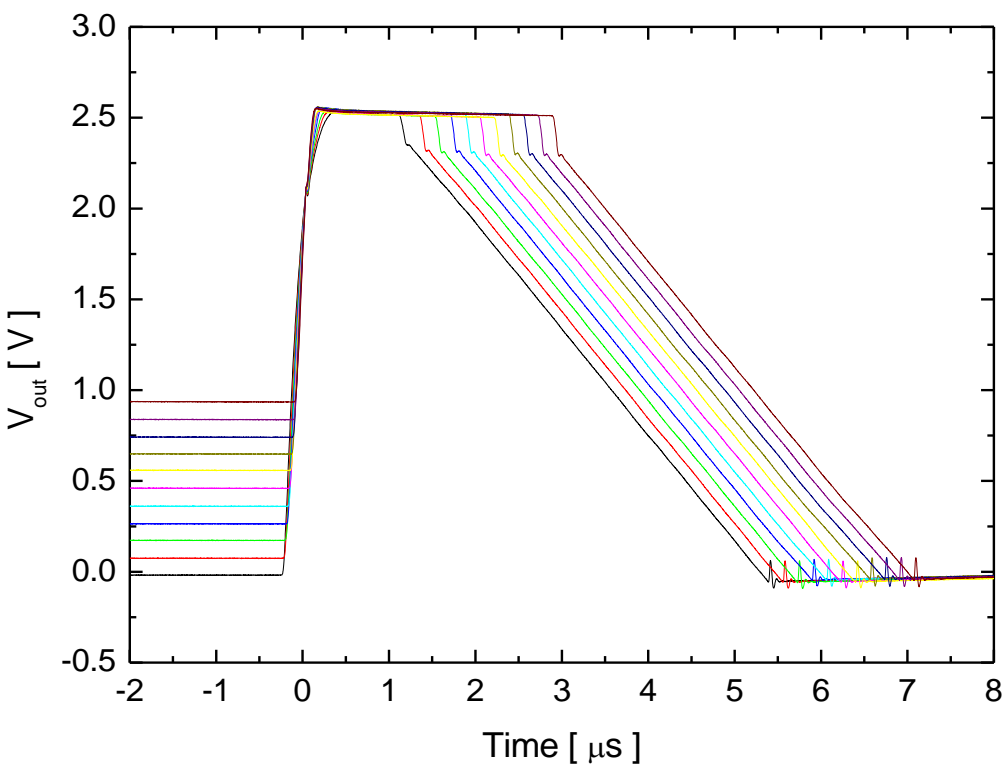
# Active Stoppers for Experimental tests with detector (previous chip version)



**Silicon detector 1 mm thick, 32 active channels, Am-Cu-Pu mixed alpha source**

# Low-pass filter





**A signal risetime with time constant comparable to the one of the LPF (or higher) induces some errors in the rejection algorithm**